

Article

1 MW-Class High-Speed Interior Permanent Magnet Synchronous Machines for Electrical Aviation Propulsion

Yang Xiao ^{1,*}, Xingqi Lyu ¹, Jinning Zhang ¹, Anshan Yu ², Yinzhaoh Zheng ² and Ruichi Wang ³

¹ School of Engineering, University of Leicester, Leicester LE1 7RH, UK; xl430@leicester.ac.uk (X.L.); jz388@leicester.ac.uk (J.Z.)

² School of Electrical and Electronic Engineering, University of Sheffield, Sheffield S1 3JD, UK; yuanshan568999@gmail.com (A.Y.); yinzhaoh.zheng@sheffield.ac.uk (Y.Z.)

³ Wolfson School of Mechanical, Electrical and Manufacturing Engineering, Loughborough University, Loughborough LE11 3TU, UK; r.wang@lboro.ac.uk

* Correspondence: yx224@leicester.ac.uk

Abstract

This paper investigates the feasibility of an interior permanent magnet (IPM) rotor for 1 MW-class high-speed permanent magnet synchronous machines (PMSMs) in a hybrid propulsion system of electrified aviation. A double-layer IPM machine and a surface-mounted PM (SPM) benchmark machine with Halbach-array PMs, which are typically employed in aviation applications; are designed using the same design specifications, the same stator, double-three-phase winding layout, physical air-gap length, outer and inner diameters of rotor; and the same materials. The rotor robustness of the IPM machine using high-strength iron material has been verified through mechanical strength analysis with an outstanding safety factor margin. The electromagnetic performances of IPM and SPM benchmark machines are compared. It is found that the IPM design can achieve similar high torque/power density and high efficiency to the SPM benchmark machine, using 48% less rare-earth PM materials and a simpler rotor structure without a carbon fiber sleeve for easy manufacturing. The investigation confirms the feasibility of IPM topology for MW-class high-speed aviation propulsion machines for lower cost and more sustainable purposes.

Keywords: interior permanent magnet rotor; permanent magnet synchronous machine; high-speed machine; hybrid propulsion

1. Introduction

The global aviation sector is currently navigating a profound technological transformation from full fossil fuel power towards electrification, which is a movement primarily spurred by the urgent demands to mitigate the environmental impact and emission of air travel, and to enhance operational efficiency [1,2]. The critical shift towards cleaner and more sustainable propulsion technologies in the aviation industry addresses air travel's role as a significant contributor to global carbon dioxide and nitrogen oxides emissions that damage the ozone layer. This issue calls for innovation and advancement in full or partial electrification through the integration of on-board power electronics, electrical machines and power distribution systems [3,4]. In the literature, different concepts and architectures of electrical propulsion systems have been proposed, including all-electric aircrafts (AEA) with full power from energy storage in the form of electricity, and more electrical aircrafts (MEA) that utilize the concepts of hybrid electrical or turbo-electrical power propulsions to enhance the efficiency of fuel-based turbine systems with reduced



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emissions [1,5,6]. Although AEA is more promising for the ambitious goal of a net-zero aviation industry, the current capability of electrical energy storage density in battery packs limits the applications of the AEA concept to commercial flights using single-aisle or double-aisle airliners. Thus, the MEA concept with hybrid electrical propulsion has gained increasing attention as the solution for electrical propulsion [7], with the main concepts including parallel hybrid electrical propulsion, all-turboelectric propulsion, series hybrid electrical propulsion, partially turboelectric propulsion, and series/parallel partial hybrid electric propulsion. A typical schematic concept of hybrid electrical propulsion system is briefly shown in Figure 1.

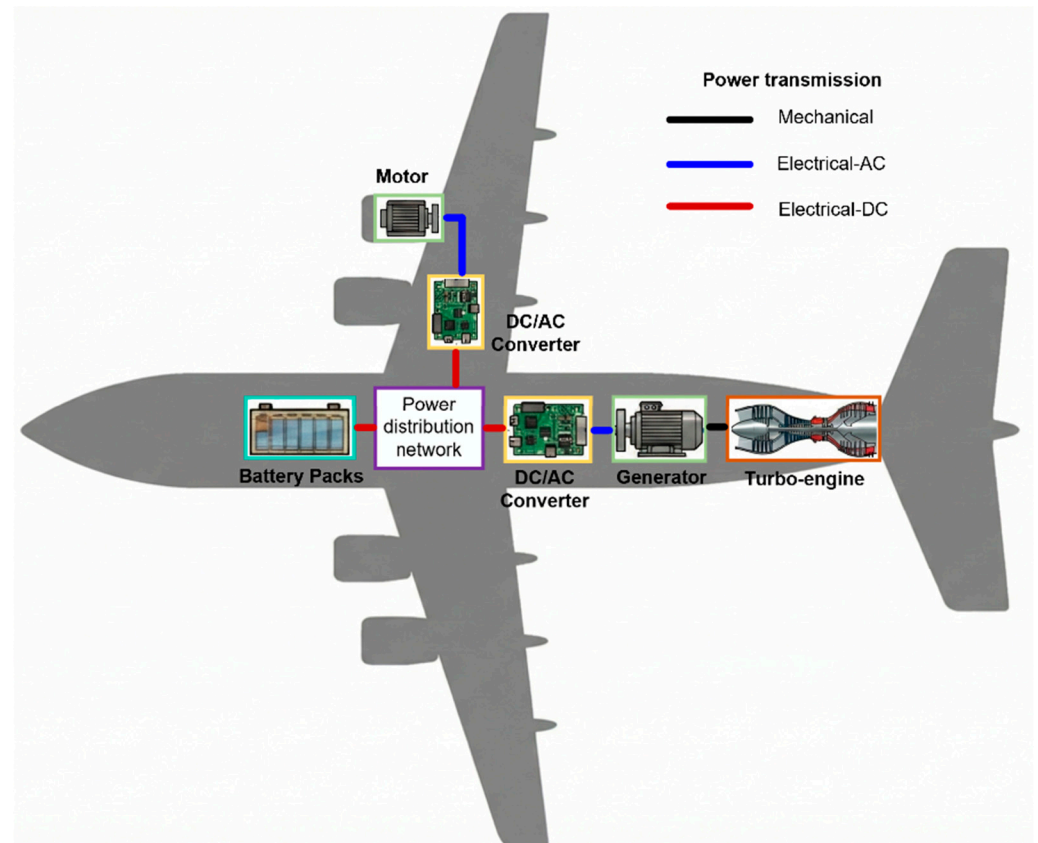


Figure 1. Typical schematic concept of hybrid electrical propulsion system in electrical aviation.

The transition to hybrid electric propulsion fundamentally alters the aircraft's power system architecture, replacing traditional hydraulic and pneumatic power sources with electrical counterparts [8]. This shift results in a dramatic increase in on-board electrical power demand, moving from hundreds of kilowatts in conventional aircraft to the multi-megawatt (MW) scale required for propulsion and high-power generation in regional and single-aisle aircraft [9]. Future projections indicate that the power capacity on aircraft could exceed ten MW [10]. To meet this unprecedented demand, the core challenge lies in developing electrical machines that possess an extremely high specific power density, high efficiency, mechanical strength, and exceptional reliability, all while operating under the severe constraints of the aerospace environment, including high altitude, low pressure, and wide temperature variations [11]. As a result, the design of electrical machines must consider multi-physics characteristics at multiple domains including electromagnetic, mechanical, and thermal performance.

Although different types of electrical machines have been considered for aerospace applications including wound-field synchronous machines (WFSMs), induction machines (IMs), and switched reluctance machines (SRMs) [12], permanent magnet (PM) synchronous

machines have emerged as the leading candidate thanks to the superior torque/power density, light weight, high efficiency, and fault-tolerant capability [13], as the main competitors show major drawbacks of lower efficiency and reliability concerns (WFSM), a lower power density and power factor (IM), and higher noises and vibrations (SRM).

There are various topologies of PM machines that have been considered and investigated for aviation applications. Axial-flux PM (AFPM) machines, which offer advantages in terms of compactness and potentially high torque density, have been investigated for 3–5 k r/min speed and 100–700 kW applications [14,15]. The industrial demonstrations confirm that the yokeless AFPM concepts can increase power density by reducing the iron mass and shortening the end windings. However, the geometrical characteristics of existing AFPM machines with short axial length and large rotor diameter restrict the specifications of high speed and high electrical loading due to concerns regarding mechanical and thermal reliability [16], limiting the applications towards MW-class high-speed propulsion cases. The slotless stator concept with air-gap windings can reduce core losses and mitigate torque ripple and cogging torque, which has been employed for a 2.5 MW aviation generator design in [17] and shows 99% efficiency. Double-rotor and double-stator topologies have been explored in [18] through comparison with conventional single-rotor concepts using both outer and inner rotor designs. It is found that the double-rotor design shows superior power density reaching 34 kW/kg at 18,000 r/min, around 50% higher than the second highest single-inner-rotor design. Nevertheless, double-rotor concept is mechanically complicated particularly from the manufacturing perspective, and the structure will also significantly limit the cooling capability of the stator, which restricted the practical application of double-rotor approaches.

In the literature, a single-rotor PM machine with surface-mounted PMs (SPMs) and slotted stator remains the dominant selection for MW-class high-speed PM machines for electrical aviation application, which are briefly summarized in Table 1. In general, most of the cases use inner-rotor topologies and Litz wire windings to reduce AC copper loss, while the out-rotor concept is investigated as a potential approach to integrate the machine into ducted turbo-fan for a compact solution [19]. An attempt to utilize hairpin windings in the stator armature is reported in [20], which aims to rebalance the DC and AC copper losses while reducing the overall volume and weight. In general, it is found, particularly by results from experimental demonstrations in [19–23], that MW-class PM machines can reach >11 Nm/kg high torque density and >97% high efficiency at the rated full power condition.

Table 1. Summary of state-of-the-art MW-class PM machines.

| Rotor Topology | Winding Types | Speed (kr/min) | Power (MW) | Active Power Density (kW/kg) | Active Torque Density (Nm/kg) | Efficiency (%) | Reference |
|-----------------|-----------------|----------------|------------|------------------------------|-------------------------------|----------------|-----------|
| Out-rotor SPM | Litz wire | 12.5 | 1 | 17 | 13.0 | 97.3 | [19] |
| Out-rotor SPM | Litz wire | 18 | 1 | 14 | 7.43 | >97 | [21] |
| Inner-rotor SPM | Litz wire | 20 | 1 | 23.6 | 11.3 | 96.9 | [22] |
| Inner-rotor SPM | Litz wire | 15 | 2.5 | 24.4 | 15.5 | >99 | [10] |
| Inner-rotor SPM | Litz wire | 15 | 4 | 17.3 | 11.0 | >97 | [8] |
| Inner-rotor SPM | Hairpin winding | 10.8 | 1 | 38.5 | 34.0 | 98.5 | [20] |
| Inner-rotor SPM | Litz wire | 20 | 1 | 23.7 | 11.3 | 97.3 | [23] |

All existing designs shown in Table 1 utilize Halbach array thanks to the superior characteristics of the air-gap side-enhanced magnetic field that reduce rotor yoke thickness and sinusoidal magnetic field distribution, thereby improving power density by further reducing the weight of the rotor iron core and improving efficiency through less magnetic harmonic distortion, respectively. However, the drawbacks of Halbach-array-based PM arrangement design include significant rare-earth PM usage with high cost and complexity of manufacturing, particularly of PM magnetization, bonding of many PM segments, and sleeve assembly [9,16,19–23].

Interior PM (IPM) machines with PMs embedded inside the rotor iron core have been widely used in electrical vehicle industries [24,25], which benefit from simple and easy manufactured rotor structure and less expensive rare-earth PM usage to ease supply-chain concerns and enable more cost-effective solutions compared with SPM designs. However, one of the most critical concerns for employing IPM topology in high-power-density and high-speed PM machines in aviation applications is the mechanical strength, as regular silicon steel material with 0.2–0.5 mm thickness usually has a yield strength at around 300–400 MPa, which usually fails to ensure a reasonable safety margin for the reliability and rotor robustness of aviation applications. The development and commercialization of high-strength silicon steel, which shows the increase in yield strength to be around 1000 MPa [26] and originally targeted the EV sector, opens a new opportunity for applicable IPM solutions in the electrical aviation sector.

As all existing papers only focus on SPM rotor topologies for MW-class high-speed PM synchronous machines for electrical aviation propulsion applications, this paper aims to fill the gap by exploring the feasibility of a 1 MW, 12 k r/min IPM machine for a hybrid electrical propulsion system in an electrical aircraft, using it as a generator. The advantages and disadvantages of IPM rotor-based MW-class high-speed PM machines are revealed by comparison with an SPM machine using Halbach-array PMs, with the same specifications, stator design, physical air-gap length, and rotor inner diameter. It is firstly validated that an IPM machine can achieve similar power density for active material weight and similar efficiency considering mechanical robustness, with much lower rare-earth PM material usage, lower cost, and easier manufacturing process.

This paper is organized as follows. In Section 2, the specification of 1 MW generator for hybrid aviation propulsion and the selections of topologies and designs of both the IPM machine and the SPM benchmark machine using Halbach-array PM configuration are given. Section 3 presents the mechanical stress analysis of the IPM machine design to validate the mechanical robustness of the IPM rotor. The electromagnetic performance of IPM and SPM machines are compared in Section 4. Finally, a conclusion is given in Section 5.

2. Machine Specification and Designs

In this paper, an IPM machine and a SPM benchmark machines using Halbach-array PMs are designed using the same specifications for a 1 MW-class, high-speed generator in the hybrid electrical propulsion system of MEA. The rated power of the generator has been designed to 1.05 MW, which is also the peak power delivered from the integrated generation system with a turbo-engine and generator that are mainly used during the take-off condition. To enable a fair comparison, both IPM and SPM machines are designed using the same stator, armature winding layout, physical air-gap-length, and rotor inner diameter to ensure the same volume of machine. To improve power density, reduce armature-harmonics-related losses, and enable fault-tolerant capability, dual-three-phase (DTP) stator winding layout will be employed, similar to other reported 1 MW-class aviation generator demonstrators [19,20]. For the SPM machine, a retaining sleeve made of

non-conductive and non-permeable carbon fiber is employed to provide pre-tightening force for Halbach-array PMs with PM segments, securing the mechanical strengths of SPM rotor at high-speed conditions. Although thinner sleeves can also be used in IPM machines for mechanical robustness enhancement, such as the application in Tesla EV motors, it has not been employed in the IPM rotor design in this work, which means the IPM machines only use rotor iron core with embedded PMs.

In this section, the generator specifications that are applied to both IPM and SPM benchmark machines and the same stator design shared by SPM and IPM machines are given first. Then, the rotor designs of both the SPM benchmark machine using Halbach-array PMs and the IPM machine are presented using the same physical air-gap length and rotor inner diameter.

2.1. Specifications of 1 MW-Class High-Speed Generator and Stator Design

As reviewed in Section 1, the primary goals for electrical machine design in the aviation industry focus on enhancing the power density as the ratio between rated power and weight of active materials and efficiency. The specifications of a 1 MW-class high-speed generator for the hybrid electrical propulsion system in MEA are given in Table 2, which shows similar characteristics of size, rated power, efficiency, speed, DC link voltage, current density, and thermal constraints as other demonstrations reported in the literature [10,17,19–23]. The criteria of mechanical rotor safety factor (SF) is defined as the ratio between maximum stress at the rated condition and the yield/tensile strengths of rotor materials, as given in (1), where σ_{max} is maximum von-Mises stress (for iron core) or Maximum Principal stress (for PMs), and σ_y is the yield strength for iron materials or tensile strength for PMs.

$$SF = \frac{\sigma_{max}}{\sigma_y} \tag{1}$$

Table 2. Main specifications of hybrid electrical propulsion generator.

| Parameters | Values |
|---|------------------|
| Rated power | 1.05 MW |
| Rated speed | 12,000 r/min |
| Rated torque | 835 Nm |
| DC bus voltage | 1 kV |
| Stator winding hotspot temperature | <180 °C |
| Rotor PM hotspot temperature | <100 °C |
| Number of phases | Dual three phase |
| Number of poles | 16 |
| Stator outer diameter | 300 mm |
| Axial length | <300 mm |
| Active power density | >17 kw/kg |
| Efficiency | >97.5% |
| Mechanical rotor safety factor at rated condition | >2.0 |

The stator is designed based on the specifications and requirements of high-speed aviation propulsion generators, and the main stator design parameters are listed in Table 3. High-permeability Fe-Co-V alloy material with higher saturated magnetic flux density compared with conventional electrical silicon steel core material is employed for the stator core to improve the power density of machine, which is usually used for machines in the aviation sector that requires superior power density. Litz bundles are used for stator coils to minimize AC winding considering the high-frequency currents flow through the windings

and high-frequency magnetic fields, including fundamental and harmonic components, that interact with the conductive coils, which may generate significant AC copper loss due to the skin effect and proximity effect in high-speed machines using form-wound copper wires or hair-pin coils. The geometry of the stator core and winding layout in one pole are shown in Figure 2a, and the phase EMF phasor diagram is illustrated in Figure 2b. The same stator design will be shared by both IPM and SPM benchmark machines to indicate a fair comparison. The drive circuit of the DTP PM machine is illustrated in Figure 3, as the two sets of three-phase windings are controlled individually by two three-phase inverters to offer redundancy and fault-tolerant capability. The power electronic DC/AC inverter using a verified Si IGBT-based power module is employed with 20 kHz switching frequency to consider the influences of harmonic currents on the machine's performance [27].

Table 3. Main design parameters of stator.

| Parameters | Values |
|--------------------------------------|----------------------|
| Slot number | 96 |
| Pole number | 16 |
| Stator outer diameter | 300 mm |
| Stator inner diameter | 246 mm |
| Slot type | In-parallel slot |
| Slot width | 4 mm |
| Slot depth | 19 mm |
| Conductor number per slot | 2 |
| Coil pitch | Full-pitch |
| Parallel branches per phase | 8 |
| Winding connection | Delta |
| Stator stack length | 190 mm |
| Stator length including end-windings | 250 mm |
| Stator core material | Fe-Co-V alloy |
| Stator core plate thickness | 0.1 mm |
| Stator winding material | Cu |
| Stator winding wire type | Litz wire |
| Maximum current density | 20 A/mm ² |
| PWM switching frequency | 20 kHz |

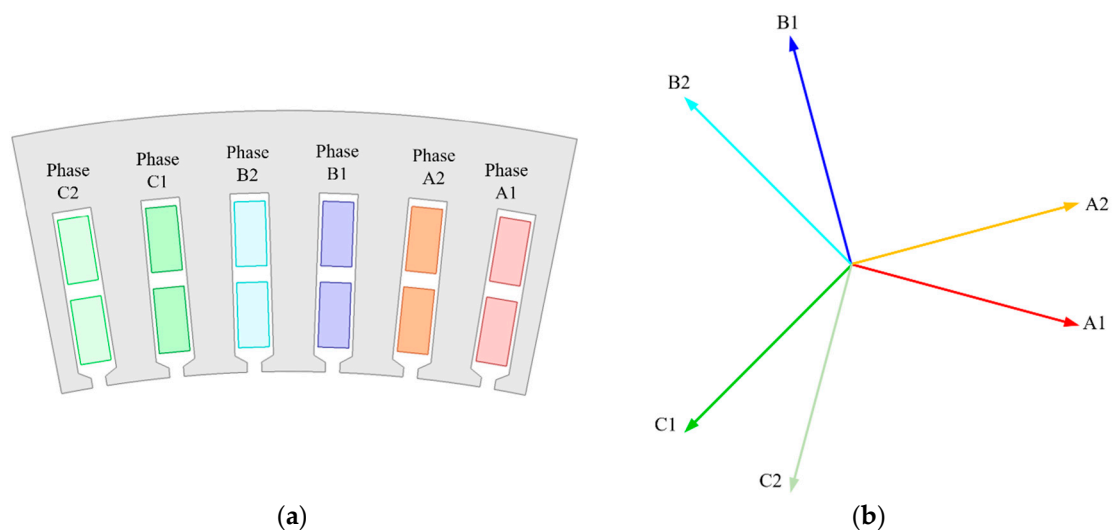


Figure 2. Stator design of 1 MW PM machine using DTP winding layout. (a) Stator core geometry and winding layout. (b) Phase EMF phasor diagram.

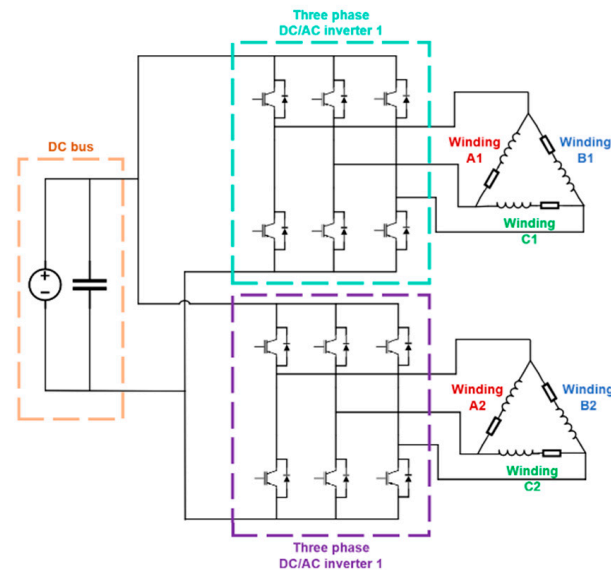


Figure 3. Drive circuit of the DTP PM machine.

2.2. Rotor Designs of IPM and SPM Benchmark Machines

The SPM benchmark rotor and IPM rotor are designed and optimized using the same outer and inner diameters, PM material and iron core material. The SmCo PMs are employed as a regular selection in the aviation sector due to outstanding thermal stability. Due to the same stator inner diameter and the same rotor outer diameter, the physical air-gap of both IPM and IPM machines remains the same. The geometries of optimized IPM and SPM benchmark machines are illustrated in Figure 4, where blue arrows are used to denote the magnetization directions of PMs. The geometrical parameters denoted in Figure 4 are mainly focused on rotor design parameters, including the outer and inner diameters of rotor and geometrical parameters of PMs. The main design parameters of both machines, including geometrical parameters denoted in Figure 4, are listed in Table 4.

The IPM rotor is designed using the same air-gap length δ and outer and inner diameters of rotor, R_{out} and R_{in} , respectively. Thus, the IPM rotor has the same volume as the SPM rotor including the sleeve. As can be seen in Figure 4, a double-layer V+ U shape IPM topology is selected to better utilize both PM and reluctance torques after comparison between regular IPM topologies, in order to improve the torque density and reduce the harmonics of rotor magnetic field and back EMFs. The iron bridges and ribs are designed to ensure the mechanical strength with an outstanding safety margin as required in Table 2. As shown in Table 4, the PM usage in the SPM benchmark machine is significantly larger than the IPM machine, with 94.3% more PM volume, which demonstrates the merit of IPM machines in reducing rare-earth PM usage.

The rotor design of SPM machine employs Halbach-array using 4 segments of PMs per pole [10], with sinusoidal magnetization to ensure a sinusoidal magnetic field with high torque density and low torque ripple while also considering a compromise between performance and manufacturability. A carbon fiber is employed to protect the PMs and secure the mechanical strength of rotor, as many components of PMs are attached at the surface of a thin rotor back iron, which is also an advantage of Halbach-array PMs by enhancing the magnetic field in the air-gap side while suppressing it in the rotor yoke side to reduce the thickness and yoke of rotor iron. To mitigate the torque ripples and harmonics, an axially V-shaped 5 segment stepping design is also included in the IPM rotor.

Table 4. Main design parameters of IPM and SPM rotors.

| Parameters | Values |
|--------------------------------------|---------------------------|
| Physical air-gap length δ | 1.5 mm |
| Rotor outer radius R_{out} | 121.5 mm |
| Rotor inner radius R_{in} | 104 mm |
| PM material | SmCo Recoma 35E [28] |
| Remanence | 1.19 T |
| Iron core material | 35SWY900 [26] |
| Sleeve material (SPM only) | Carbon fiber |
| IPM | |
| Magnet layer | 2 |
| PM1 width W_{pm1} | 7.8 mm |
| PM1 height h_{pm1} | 4 mm |
| PM2 width W_{pm2} | 14.8 mm |
| PM2 height h_{pm2} | 6 mm |
| PM3 width W_{pm3} | 3.5 mm |
| PM3 height h_{pm3} | 4.5 mm |
| Volume of PMs | 825,360 mm ³ |
| Axial segments of PMs | 3 |
| Rotor axial segments | 5 |
| Type of axial stepping | V-shape |
| Mechanical angle between steps | 0.75 degree |
| SPM benchmark | |
| Sleeve thickness h_s | 2 mm |
| PM height h_{pm} | 12 mm |
| Rotor back iron thickness h_{iron} | 3.5 mm |
| Axial segments of PMs | 5 |
| Volume of PMs | 1,604,474 mm ³ |

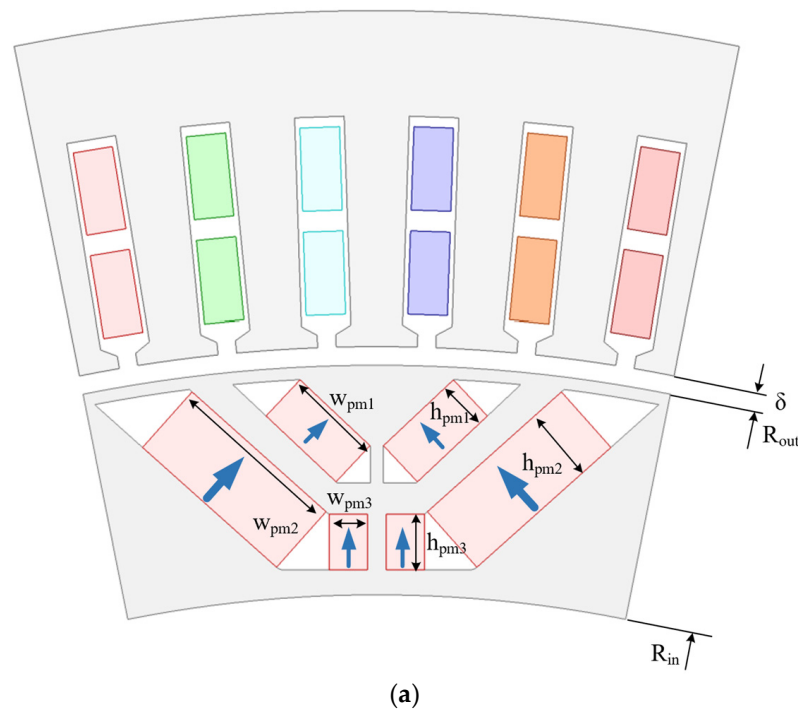


Figure 4. Cont.

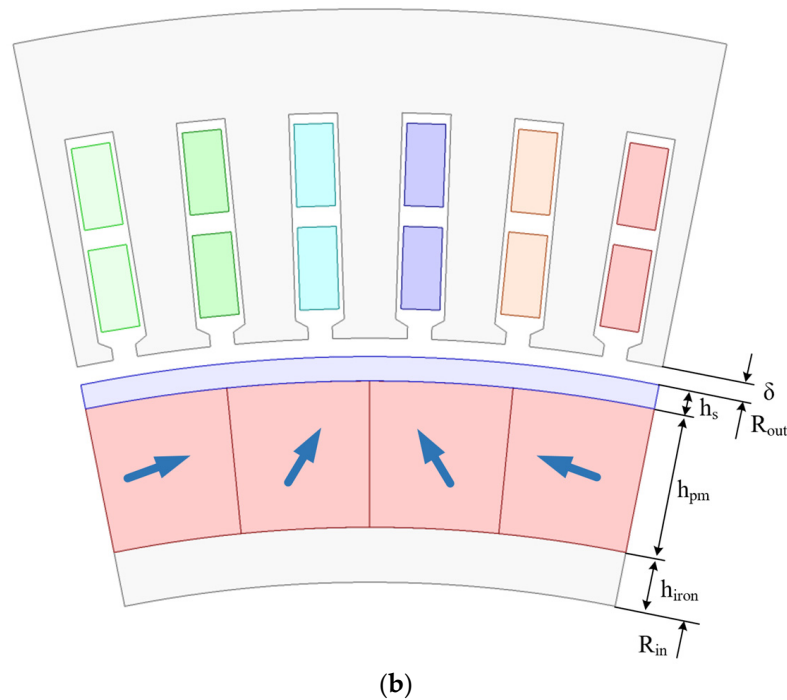


Figure 4. Rotor geometries and key design parameters of 1 MW IPM and SPM benchmark machines. (a) IPM. (b) SPM.

3. Mechanical Strength Analysis for IPM Machine Design

One of the main concerns for the feasibility of utilizing IPM topologies in MW-class high-speed PM machines in the electrified aviation sector remains to be the mechanical robustness and notable safety margin criteria, as the mechanical strength of a SPM machine design with Halbach-array and a carbon fiber sleeve with similar size, speed, and design parameters has been validated in the literature, and reported demonstrations are given in Table 1. In this section, mechanical finite element (FE) analysis is carried out to calculate the Von-Mises stress distribution on the rotor iron core and the maximum principal stress on the PMs, using PrePoMax mechanical analysis platform. Through comparison with yield stress for iron core material and tensile stress for PM materials, respectively, the mechanical safety factors across a speed range, particularly on the rated speed, can be obtained for iron core and PMs to validate the mechanical strengths of the IPM design. The mechanical characteristics of iron core and PM materials used in the FE simulation are shown in Table 5.

Table 5. Mechanical properties of rotor core and PM materials.

| Item | 35SWY900 | SmCo Recoma 35E |
|-------------------------------|----------|-----------------|
| Density (kg/mm ³) | 7600 | 8300 |
| Young’s modulus (GPa) | 180 | 140 |
| Poisson ratio (-) | 0.3 | 0.34 |
| Tensile strength (MPa) | - | 35 |
| Yield strength (MPa) | 960 | - |

The distributions of Von-Mises stress on the rotor iron core at different rotation speeds and constant rated 835 Nm torque are shown in Figure 5. As can be seen, maximum Von-Mises stress remains located on the bottom of the second-layer rib across a range of speeds, reflecting the influences of both centrifugal force and torque, which is also gradually increased with the increase in rotational speed. The maximum Von-Mises stress on the rated speed and torque, namely 12,000 r/min and 835 Nm, is around 383.2 MPa,

which is significantly lower than the 960 MPa yield strength of used rotor iron steel material, demonstrating the mechanical robustness of IPM design at the rated condition. The distributions of maximum principal stress on the PMs at different rotation speeds and constant rated 835 Nm torque are shown in Figure 6. As illustrated in Figure 6a, the maximum principal stress on the PMs reaches around 16.66 MPa, which is smaller than half of the tensile strength of the PMs, also demonstrating the mechanical strength of the PMs at the investigated conditions.

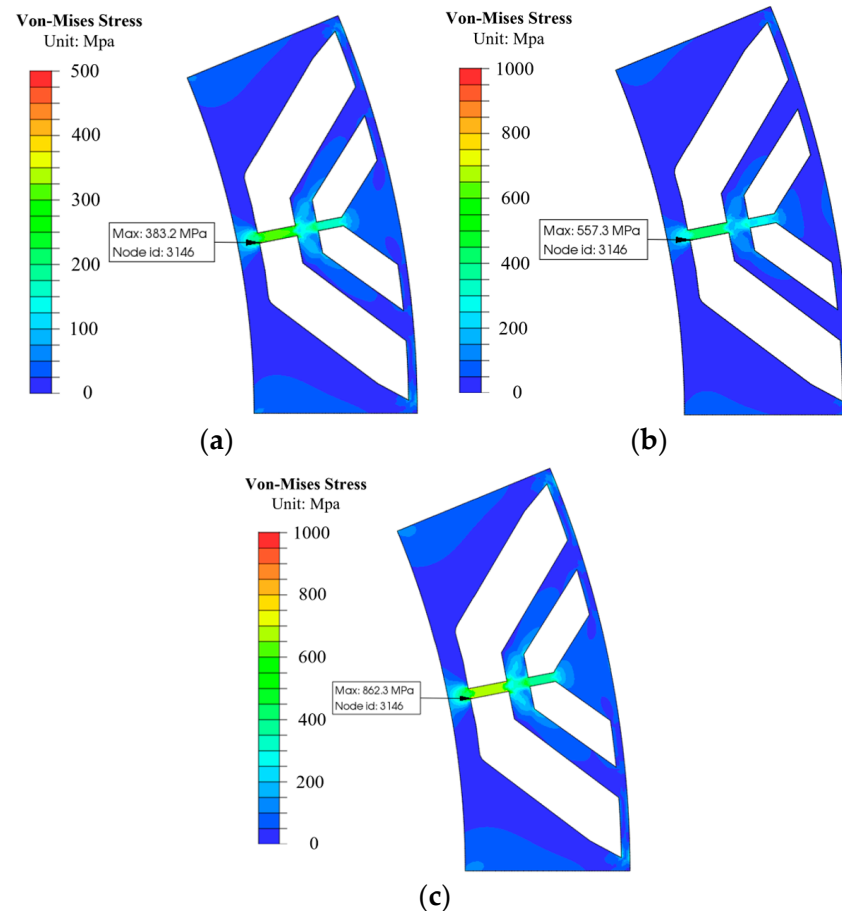


Figure 5. Von-Mises stress distributions in rotor core at different rotation speed and constant rated 835 Nm torque. (a) 12,000 r/min. (b) 15,000 r/min. (c) 18,000 r/min.

To better understand the safety margin of mechanical strengths in the IPM designs across a speed range for both iron core and PMs, the safety factors of rotor core and PMs are calculated for different speed conditions based on Equation (1), as shown in Table 6. It confirms that the mechanical safety factor at the rated condition meets the criteria of $SF > 2$ given in Table 1, which validates the mechanical robustness of the proposed IPM design. Moreover, the comparison also indicated that the maximum speed of this IPM design should not exceed 16,500 r/min, considering a minimum >1.1 safety factor as a boundary condition. As the maximum stress on the iron core reaches the criteria at close to 13,500 r/min, it may be possible to further increase the rated speed of this design for the purpose of power density enhancement with the same output torque, subject to the careful improvement on PMs' geometric and assembly designs to ensure an acceptable SF on PMs.

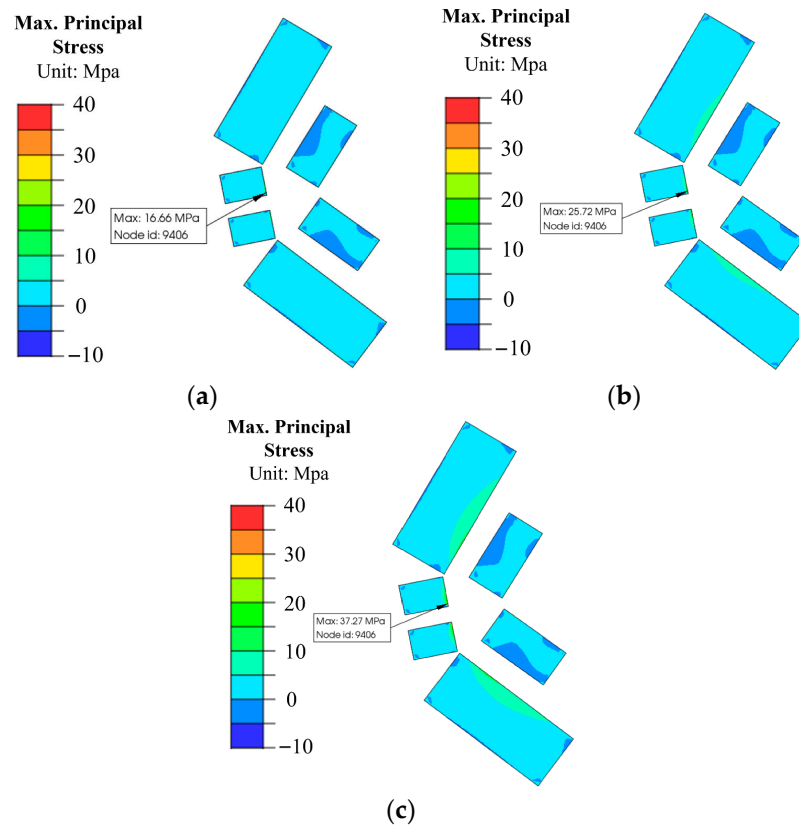


Figure 6. Maximum principal stress distributions on PMs at different rotation speed and constant rated 835 Nm torque. (a) 12,000 r/min. (b) 15,000 r/min. (c) 18,000 r/min.

Table 6. Safety margin of mechanical strength for IPM machine across a speed range.

| Speed (r/min) | Rotor Iron Core | | PMs | |
|---------------|-----------------------------|--------|-----------------------------|--------|
| | Max. Von-Mises Stress (MPa) | SF (-) | Max. Principal Stress (MPa) | SF (-) |
| 12,000 | 383.2 | 2.51 | 16.66 | 2.10 |
| 13,500 | 481.5 | 1.99 | 21.15 | 1.65 |
| 15,000 | 557.3 | 1.72 | 25.72 | 1.36 |
| 16,500 | 698.9 | 1.37 | 31.51 | 1.11 |
| 18,000 | 852.3 | 1.13 | 37.27 | 0.94 |

4. Comparison of Electromagnetic Performance Between IPM and SPM Benchmark Machines

While the analysis in Section 3 confirms the mechanical strengths and feasibility of IPM machine design for the design specifications and criteria of a hybrid aviation propulsion generator, this section will focus on the comparison of electromagnetic performance between IPM and SPM machines to understand the merits and demerits of IPM machines compared with the SPM benchmark machine whose topology has already been widely employed in the electrical aviation industry, using the Ansys Electronics 2024 and MotorXP-PM electromagnetic FE analysis platforms.

4.1. Open-Circuit Performance

The distributions of magnetic flux density and flux lines at open-circuit conditions in IPM and SPM benchmark machines are illustrated in Figure 7. As can be seen, the SPM benchmark machine has higher magnetic flux density in the stator side, due to more PM usage, the superior magnetic enhancement effect of the Halbach-array, and notable leakage

fluxes in the rotor bridges and ribs of IPM machine, while the magnetic flux density in the SPM machine has also reduced due to the influences of non-permeable carbon fiber sleeve with 2 mm thickness, which actually increase the equivalent air-gap length of the SPM machine to 3.5 mm, compared with the 1.5 mm equivalent air-gap length in the IPM machine that remains the same as the physical air-gap length.

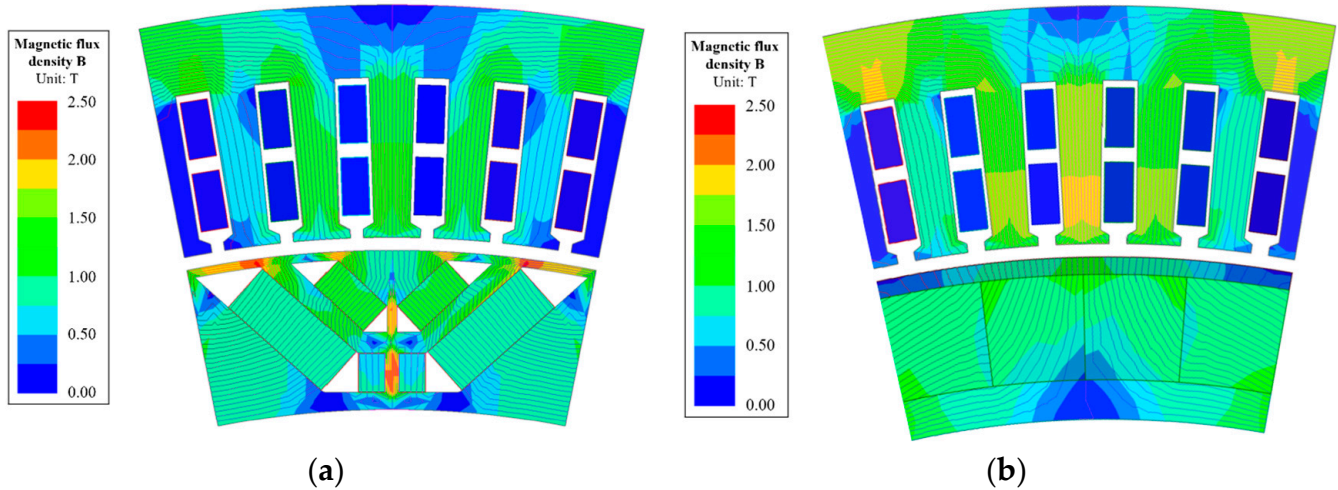


Figure 7. Distributions of magnetic flux density and flux lines at open-circuit condition. (a) IPM (b) SPM.

Consequently, the SPM benchmark machine shows about 15% higher fundamental flux density as shown in Figure 8a and 15% higher fundamental back electromotive force (EMF) as shown in Figure 9a, which matches the observed magnetic flux density distributions in Figure 7 and aligns with the better magnetic focusing effect of Halbach-array PMs and more PM usage. However, in the perspective of magnetic flux density and back-EMF per PM volume, the IPM machine shows much better performance than the SPM benchmark machine, as the IPM machine shows 60% higher open-circuit magnetic flux density per used PM volume, which demonstrates outstanding PM usage ratio for the IPM machine.

In addition, the SPM benchmark machine also shows significantly lower harmonic distortions, as evidenced by more than 50% and 59% lower total harmonic distortions (THDs) for magnetic flux density and back EMFs, respectively, compared with that of the IPM machine. It also reflects the outstanding harmonic elimination effect of Halbach-array PMs, even though the rotor stepping design with harmonic suppression effect has already been employed in IPM machines.

Due to the nature of IPM and SPM topologies, and further influenced by the longer equivalent air-gap length in the SPM design, the IPM machine shows higher *d*- and *q*-axis inductances compared with that of the SPM machine. In addition, the IPM machine shows a saliency ratio of 1.75, as reluctance torque can be utilized to enhance the torque/power density, as shown in Table 7.

Table 7. Comparison of open-circuit inductances between IPM and SPM machines.

| Inductance | IPM | SPM |
|-------------------------------|------|------|
| d-axis inductance (mH) | 0.12 | 0.08 |
| q-axis inductance (mH) | 0.21 | 0.08 |
| Saliency ration L_q/L_d (-) | 1.75 | 1.0 |

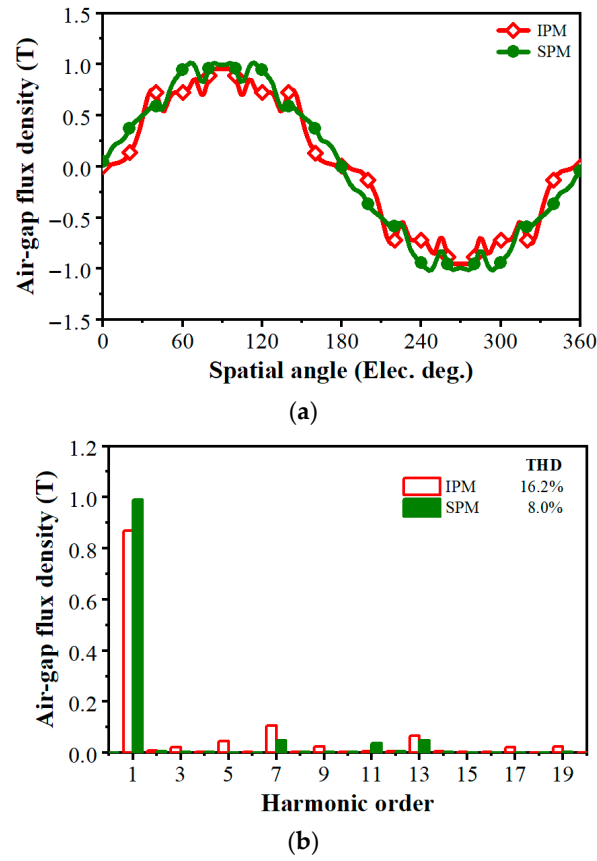


Figure 8. Comparison of open-circuit air-gap magnetic flux density. (a) Waveform (b) Spectrum.

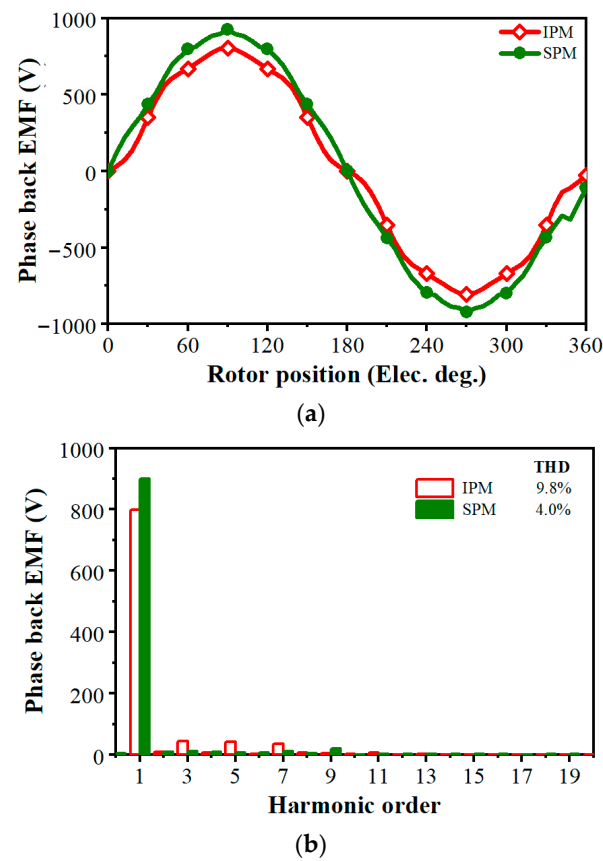


Figure 9. Comparison of back-EMFs. (a) Waveform (b) Spectrum.

4.2. On-Load Performance, Power Density, and Efficiency at Rated Condition

The on-load performance of both IPM and SPM benchmark machines are analyzed at rated 12,000 r/min speed, 1.05 MW output power conditions, and compared to verify the feasibility of employing IPM machine concept for 1 MW-class high speed propulsion generator application in hybrid aviation propulsion system, with particular focus on power density and efficiency.

The torque current angle characteristics of IPM and SPM machines are shown in Figure 10. As can be seen, the maximum torque of the SPM benchmark machine is obtained when the current advancing angle is 0, namely d-axis current $i_d = 0$, due to ignorable reluctance torque. Thus, the $i_d = 0$ control can be applied for maximum torque/power density. For IPM machines, the maximum torque is achieved at a larger current advancing angle due to utilization of both PM torque and reluctance torque components to enhance torque/power density without more PM usage. As extracted in Figure 11 using the frozen permeability method [29], both PM and reluctance torque components contribute to the maximum torque, with about 12% and 88% average torque from reluctance torque and PM torque components, respectively.

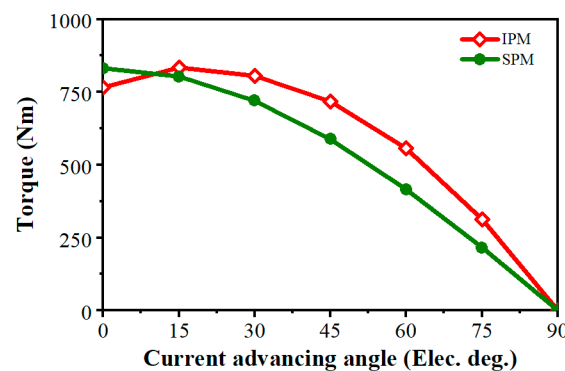


Figure 10. Comparison of torque–current angle characteristics.

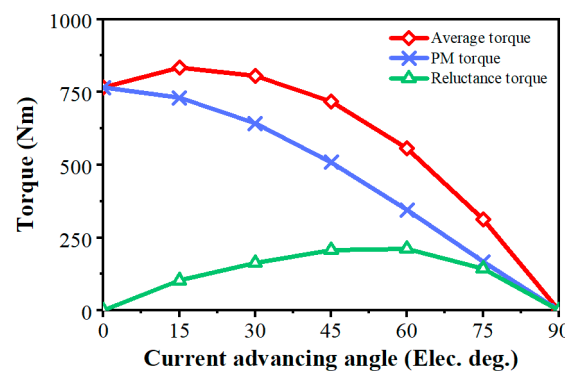


Figure 11. Extraction of torque–current angle characteristics in IPM machine.

The on-load torque waveforms of SPM and IPM machines at the rated condition are shown in Figure 12a, while their spectrums are compared in Figure 12b. As can be seen, both machines have low torque ripples (<5%), which is the percentage ratio between the peak-to-peak value of the torque waveform and the average torque, thanks to rotor stepping design in the IPM machine, albeit to a compromise of average torque, and the Halbach-array PMs in the SPM benchmark machine, respectively. The main order of torque harmonics in both IPM and SPM machines is 13th, which is originated from tooth harmonics due to slot opening. In addition, the IPM machine shows 60% further lower torque ripple compared with that of the SPM benchmark machine.

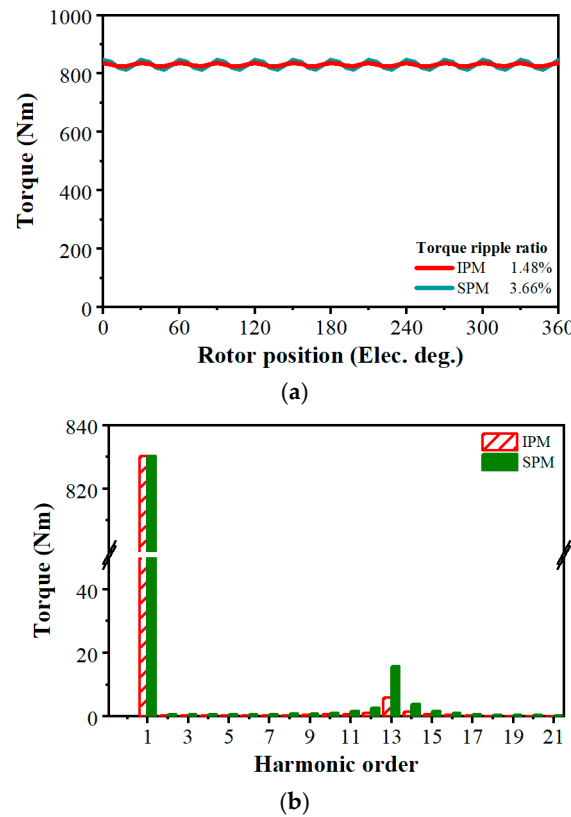


Figure 12. Comparison of on-load torque performance. (a) Waveform (b) Spectrum.

The torque-speed performance of IPM and SPM machines at the same 1 kV DC bus voltage and maximum 20 A/mm² current density are illustrated in Figure 13. It reveals that the IPM machines show wider speed range and higher output torque/power within the overspeed range beyond the rated speed (normally within 20–30% for safety consideration) due to the better flux-weakening capability of IPM topology.

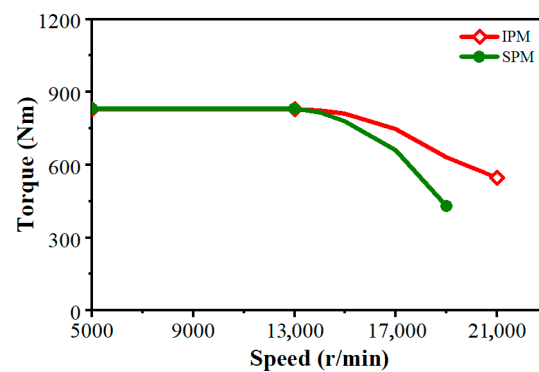


Figure 13. Comparison of torque-speed performance.

The most important performance indicators, particularly in the aviation industry, for electrical machines in the hybrid propulsion system are power density and efficiency. To evaluate the power density as the ratio between output power and total active weight, the weights of active components and the total weights as the sum of all components are given in Table 8. As can be seen, the IPM and SPM machines generally have similar total weights. In addition, the IPM shows significantly lower PM usage than the SPM benchmark machine, as the rotor PM weight in IPM is 48.6% lighter than that of the SPM machine.

Table 8. Comparison of active weights between IPM and SPM machines.

| Machine Active Components | IPM | SPM |
|---------------------------|-------|-------|
| Stator core (kg) | 25.68 | 25.68 |
| Stator windings (kg) | 6.16 | 6.16 |
| Rotor core (kg) | 11.00 | 4.39 |
| Rotor PMs (kg) | 7.31 | 14.21 |
| Rotor sleeve (kg) | - | 0.23 |
| Total Weight (kg) | 50.15 | 50.67 |

Based on the active weight obtained in Table 8, the on-load performance of IPM and SPM machines at rated conditions, with particular emphasis on power density and efficiency, are compared in Table 9. As can be seen, the current density needed for IPM machines to achieve the rated power is slightly higher than that of the SPM benchmark machine, while the current harmonic ripple in the IPM is smaller than that of the SPM due to larger inductances.

Table 9. Comparison of on-load performance at rated condition between IPM and SPM machines.

| On-Load Performance | IPM | SPM |
|--------------------------------------|--------|--------|
| Fundamental phase current RMS (A) | 252 | 246 |
| Current density (A/mm ²) | 19.1 | 18.6 |
| Current advancing angle (Elec. Deg.) | 17.4 | 0 |
| Current harmonic ripple (%) | 1.9% | 3.2% |
| Speed (r/min) | 12,000 | 12,000 |
| Torque, Power, and Power density | | |
| Output torque (Nm) | 831.5 | 830.9 |
| Output power (MW) | 1.045 | 1.044 |
| Torque ripple (%) | 1.48% | 3.66% |
| Active torque density (Nm/kg) | 16.6 | 16.4 |
| Active power density (kW/kg) | 20.8 | 20.6 |
| Losses and Efficiency | | |
| Copper loss (kW) | 7.5 | 7.35 |
| Stator Iron loss (kW) | 14.0 | 12.7 |
| Rotor Iron loss (kW) | 0.50 | 0.08 |
| Magnet loss (kW) | 0.05 | 0.72 |
| Stator side loss (kW) | 21.5 | 20.05 |
| Rotor side loss (kW) | 0.55 | 0.80 |
| Total loss (kW) | 22.05 | 20.85 |
| Efficiency (%) | 97.93 | 98.04% |
| Thermal Simulation | | |
| Stator winding hotspot temperature | 178 °C | 176 °C |
| PM hotspot temperature | 93 °C | 98 °C |

As shown in Table 9, the IPM machine can achieve similar active power density and active torque density to the SPM benchmark machine, both reaching >20 kW/kg and >16 Nm/kg, respectively, which demonstrates superior performance of both IPM and SPM benchmark machine designs in this work, as evidenced by comparable power density with the state-of-the-art demonstrations listed in Table 1. The electromagnetic losses and efficiency are also shown in Table 9. As can be seen, the IPM machine has slightly higher copper loss due to slightly larger current density applied to achieve the same output torque, and higher iron losses due to higher harmonics distortion in the magnetic flux density,

compared with the SPM benchmark machine. However, the IPM machine also shows much lower magnetic losses than SPM machine, as the PMs in IPM are embedded inside PM cavities in the rotor rather than directly exposed to all magnetic harmonics in the air-gap in the SPM benchmark machine. Consequently, the IPM machine has 5.7% higher total losses than the SPM machine, which results in a 0.11% lower efficiency of the IPM machine. As both IPM and SPM machines reach a significantly high efficiency around 98%, the difference in efficiency is minor. The thermal FE analysis has also been carried out for IPM and SPM machines, using the same cooling system design and parameters, to compare the effects of slightly different losses on the hotspot temperature distributions. The comparison confirms that both IPM and SPM machines meet the hotspot temperature criteria in Table 2.

To compare the demagnetization-withstand capability, the magnetic flux density in PMs in both SPM and IPM machines are compared at the same 350 A demagnetization phase RMS current, and the region with <0.1 T magnetic flux density is denoted as high irreversible demagnetization risk, as illustrated in Figure 14. As can be seen, at the full demagnetization current with 1.4 times the amplitude of the rated current, the IPM shows significant demagnetization-withstand capability as no regions show irreversible demagnetization risk, compared with the SPM benchmark machine that shows notable regions with irreversible demagnetization risk (<0.1 T magnetic flux density). The comparison validates better demagnetization-withstand capability in the IPM machine compared with the SPM machine benchmark.

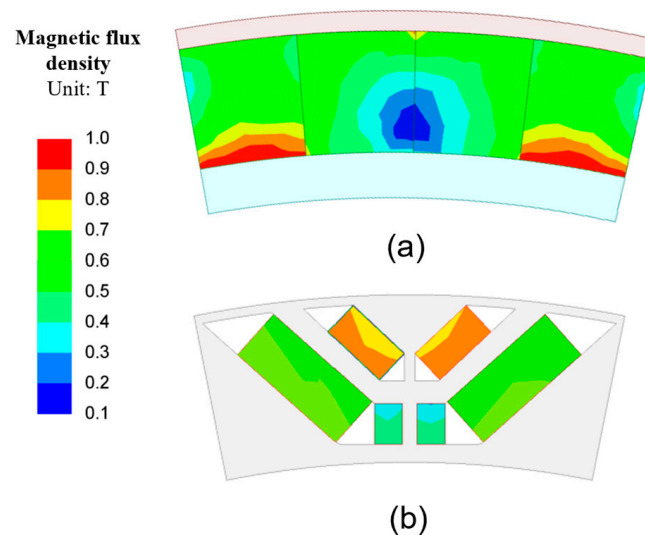


Figure 14. Comparison of demagnetization-withstand capability at 350 A demagnetization phase current RMS. (a) SPM (b) IPM.

Nevertheless, the comparison also reveals that IPM machine can achieve similar high-power density and superior efficiency to the SPM machine using Halbach-array PMs with notable mechanical strengths. Thus, it is demonstrated that IPM machine is feasible for 1 MW-class high speed propulsion machines in the electrical aviation sector. It is also confirmed that the IPM machine design shows notable merits including significantly less rare-earth PM usage, which improves sustainability, and less complicated manufacturing requirements for the rotor. These merits will also introduce the advantage of lower costs.

5. Discussions

This study investigates the feasibility of employing IPM topologies in a 1 MW-class high-speed PM synchronous generator for hybrid-electric propulsion systems in electrified aviation. Compared with the benchmark machine using SPM topologies with Halbach-

array PMs, which are the most common selection in demonstrations and in the industry at present for relevant applications, the proposed IPM design with double-layer V + U PMs configuration and high-strength iron core material 35SWY900, using the same stator, physical air-gap-length, and rotor diameters, demonstrates similar high power density, high efficiency, and adequate mechanical robustness, reflected by a significant mechanical safety margin at rated speed. It is also found that the IPM design achieves a substantial 48.5% reduction in rare-earth permanent magnet material usage and simplified rotor structure without the complicated manufacturing process of a carbon fiber sleeve, revealing potential for more sustainable and cost-effective solutions for electrical aviation propulsions.

Nevertheless, the comparison between IPM and SPM topologies reveals several important trade-offs for future practical deployment in electrified aviation systems, which will be discussed in detail in this section. From the electromagnetic perspective, the SPM benchmark machine benefits from stronger air-gap magnetic field and lower harmonic distortion due to the Halbach-array configuration and higher magnet volume, albeit with a longer effective air-gap length as the result of the rotor sleeve in SPM design. Consequently, the SPM design shows slightly lower total loss, mainly due to lower stator iron loss, and marginally higher efficiency (0.11%) are achieved. In contrast, the IPM machine exhibits higher inductances and saliency ratio, enabling reluctance torque contribution and improved flux-weakening capability. This leads to a wider operational speed range and improved demagnetization-withstand capability, which are attractive features for reducing PM usage, easing supply chain concerns of rare-earth materials, and improving the reliability of the machine.

From a mechanical and manufacturing perspective, the IPM rotor eliminates the need for a carbon fiber retaining sleeve and significantly reduces permanent magnet volume through using a simplified and mature IPM rotor assembly structure that has been well-demonstrated in EV industries, which improves manufacturability and potentially reduces production cost. The use of high-strength silicon steel enables sufficient safety margin for high-speed operation, addressing one of the major historical barriers for adopting IPM topology in MW-class aviation machines. However, the IPM design introduces increased air-gap magnetic field harmonic distortion, resulting in slightly higher iron losses due to harmonics components. These factors result in slightly higher stator-side loss and marginally lower efficiency and may impose additional thermal management requirements in the stator side, albeit to an ease of thermal management in the rotor side due to lower total rotor loss.

It should be noted that the 2 mm carbon fiber sleeve design will influence the electromagnetic performance of SPM rotors with Halbach-array PMs, as an inherent requirement of SPM rotor structure with Halbach-array PMs for high-speed machines, because the retaining sleeve is essential for the mechanical robustness of SPM rotor, particularly for Halbach-array PMs with many PM segments. In this work, the sleeve thickness has been designed to 2 mm based on the requirement of mechanical strength for a fair comparison, similar to the design employed in demonstrated references [10,19]. In addition, the modal vibration behaviors of the proposed IPM and SPM benchmark rotors are assumed to be similar, considering the same inner and rotor diameters, the same axial length, and similar rotor weights between the IPM and SPM designs in this work.

Moreover, although this feasibility study using mechanical FE analysis confirms a sufficient safety factor for the IPM design at rated conditions, experimental validation at full MW-scale prototype remains necessary before moving the concept into industry application. It is also acknowledged that a full-scale 1 MW prototype was not manufactured in this study. Instead, validated electromagnetic and mechanical FE analyses were employed to provide a fair comparison under identical design constraints. Such simulation-based evaluation

is widely adopted in early-stage feasibility studies of MW-class aviation machines, where prototype construction involves substantial cost and infrastructure requirements.

Overall, the findings confirm that the IPM topology represents a technically viable alternative to conventional Halbach-array SPM machines for MW-class aviation propulsion, particularly for improving sustainability, manufacturability and reducing costs. As this remains a feasibility study rather than an industry validation, to better reflect the advantages and disadvantages of the proposed IPM design for a 1 MW aviation propulsion machine, a SWOT analysis has been conducted to conclude the strategic advantages, limitations, and future opportunities, as shown in Table 10. Although there are some references showing designs with higher power density than the designs in this paper, the purpose of this work is not to develop a machine with highest power density on record, but is instead to compare the proposed IPM and conventional SPM machines designs using the same stator and rotor diameters to reveal the feasibility of utilizing IPM design for an alternative simple, sustainable, and cost-effective solution for MW-class aviation machines.

Table 10. SWOT analysis of proposed IPM topology for 1 MW high-speed electrical aviation propulsion machines.

| SWOT Category | Description |
|---------------|--|
| Strengths | <ul style="list-style-type: none"> • Similar high-power density and high efficiency to the conventional SPM machine using Halbach PMs; • ~48% reduction in rare-earth PM usage; • IPM rotor structure using rectangular PMs without carbon-fiber sleeve, easing manufacturing complexity; • Better demagnetization-withstand capability; • Wider flux-weakening speed range; • Low torque ripple. |
| Weaknesses | <ul style="list-style-type: none"> • Slightly higher harmonic distortion and iron loss; • Slightly higher total stator loss, leading to marginally lower efficiency than SPM (~0.1%); • Mechanical design relies on high-strength steel; • No MW-scale prototype for experimental validation yet. |
| Opportunities | <ul style="list-style-type: none"> • Growing demands for sustainable and cost-effective solutions of aviation propulsion machines; • Advances in high-saturation flux density, low iron loss, and high-strength electrical steels; • Potential for further performance enhancement through multi-physics design optimization; • Scalability to up to 10-MW propulsion systems and up to 15,000–16,000 r/min high speed applications, covering the whole market of aviation machine ranges. |
| Threats | <ul style="list-style-type: none"> • Aviation industry is generally conservative toward new techniques; • Certification requirements and reliability validation; • Manufacturing tolerance challenges at high-power density, high-speed, and high-altitude operation; • Competition from improved SPM or superconducting machine designs. |

In summary, the SWOT analysis indicates that the proposed IPM design offers significant strengths in reduced rare-earth material usage, simplified rotor structure for manufacturability, and similar power density and efficiency compared with the SPM benchmark machines using Halbach-array PMs, while its main weaknesses include higher harmonic-

related losses and the lack of full-scale experimental validation. With growing demand for sustainable and cost-effective aviation propulsion and advances in high-strength materials providing clear opportunities, challenges remain primarily in certification and validation at full MW-scale prototypes and competition with well-established SPM solutions and potential competition with superconducting machines.

6. Conclusions

In this paper, a double-layer IPM machine using high-strength iron material for rotor core and a SPM machine using Halbach-array PMs have been designed, analyzed, and compared for the 1 MW-class high-speed generator in hybrid aviation propulsion. Through electromagnetic and mechanical analyses, it is revealed that the IPM machine meets the high criteria of safety factor at rated high-speed operation conditions and shows similar >20 kW/kg high-power density and 98% high efficiency to the SPM benchmark machine, meeting the specification requirements of power density and efficiency, while using only 51.4% of total rare-earth PM usage and simplified rotor fabrication processes compared with the SPM machine. It is also found that the IPM machine shows better demagnetization-withstand capability than the SPM benchmark.

Therefore, the feasibility of using IPM topology for MW-class high-speed aviation propulsion applications has been confirmed, with highlighted less rare-earth material volume, easier manufacturing, and consequently less costs, than the conventional designs based on SPM topology using carbon fiber sleeve and Halbach-array PMs. The conclusion of this paper indicates a new technical approach for large-capacity electrical machine designs in the electrical aviation industry using IPM machines to achieve low cost and sustainability. In future work, the design solutions to ease the high harmonics distortion and consequently high iron losses in the IPM machine will be investigated, and the dynamic control strategies for torque and speed control in the IPM machine will also be explored.

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