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Theoretical Analysis of Long-Haul Systems Adopting Mode-Division Multiplexing

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14 Abstract

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16 expressions for the nonlinear interference power in birefringent few-mode fib rs (F', n') are derived and the effect of differential 17 mode group delay (DMGD) is investigated. Moreover, the nonlinearity accume and through propagation in multiple-spans fiber 18 and the birefringence effect are considered. In addition, we discuss the effect of the DMGD on the fiber nonlinearity in systems 19 20 adopting mode-division multiplexing (MDM). The results show that the DMGD n magement degrades the system performance in 21 weak coupling regime because the nonlinear interference is enhanced. How ver, frong coupling-based transmission outperforms 22 weak coupling transmission regardless of the DMGD effect in the weak coupling regime. On the other hand, by taking the DMGD 23 effect into account, the system performance in weak coupling regime in but that in strong coupling regime. Furthermore, the impact of the nonlinearity on the maximum reach is discussed.

Keywords: Few-mode fibers (FMF), Gaussian noise model (GN- toot, ...ode-division multiplexing (MDM), nonlinearity 26 modeling, space-division multiplexing (SDM).

1. Introduction

Optical transmission capacity is rapidly approaching its fun-33 damental nonlinear limit in single mode fibers (SMF) [1]. 7ptical space-division multiplexing (SDM) is a prom. ing degree of freedom that increases the fiber transmission cr pacity. The upports multiple communication channels using mod's in fewmode fibers (FMFs) and/or cores in multi-core ^{G1} ers (ACFs) [2-4]. In recent years, several experiment, efforts, we been done to demonstrate optical space-divisio . n. 'tiplexing based systems [5-7].

41 However, in long-haul transmissior, the ystem performance 42 suffers from physical impairments due to attenuation, disper-43 sion, and nonlinearity. The fiber ne alinearity is a major source 44 of capacity performance limitation [4, 8-11]. This nonlinear 45 limitation arises from the nonline.. inte action between dif-46 ferent co-propagating optical needs due to Kerr-effects. These Kerr-effects simply involve 1 onlinear changes in the refractive 48 index with increasing transmit. A signal power, thus generating 49 self-phase modulation (SPM), cross-phase modulation (XPM), 50 or four wave mixing (F VM) [12 -15]. Another linear interac-51 tion in FMFs transmissio. ari as from the coupling between 52 53

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various spatial copropagating fields that results in a periodicpower transfer from an optical field to another copropagating one [11, 16, 17]. This linear coupling can exist: between dualpolarized fields on a specific mode (core), called linear polarization coupling, or (and) between different copropagating mode (core) fields, called linear mode coupling [17]. When the linear mode coupling level is comparable to that of polarization one, two distinct coupling regimes occur, namely weak- and strongcoupling regimes. In the weak coupling regime, the linear mode coupling is insignificant and could be neglected compared to the linear polarization coupling. On the other hand, in the strong coupling regime, the linear mode coupling is significant compared to the linear polarization coupling [16]. The randomlyvarying birefringence during fiber transmission results in a reduction of the nonlinear interaction due to the randomly-averaging operation under the birefringence effect [16, 18, 19]. In the strong coupling regime, this randomly-averaging is higher compared to the weak coupling one, because of the large randomfluctuation of the propagating power in strong coupling case. Another linear propagation process in FMFs transmission is the differential mode group delay (DMGD) between the copropagating modes [20, 21]. It is similar to the differential group delay (DGD) between dual-polarized fields in SMFs transmission [22]. DMGD is a design limitation of multiple-inputmultiple-output (MIMO) receivers in MDM based systems [21, 23]. Though the DMGD leads to an increase in the complexity of MIMO-receivers, it reduces the impact of nonlinearity of FMFs based transmission [24]. Further, a DMGD-management may be performed by periodically interchanging FMFs with

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1 different DMGDs in order to reduce the complexity of MIMO 2 receivers [25].

Recently, the description of nonlinear propagation of dual-3 polarized signal through nonlinear-dispersive multi-mode opti-4 cal fibers is described by the coupled multi-mode generalized 5 nonlinear Schrödinger equation (MM-NLSE) and generalized б 7 coupled multi-mode Manakov equations [9, 16, 18, 26, 27]. 8 The generalized multi-mode coupled Manakov equations are 9 simpler than the MM-NLSE, but both have to be solved numer-10 ically by the split-step-Fourier-method (SSFM) [11, 16]. Exten-11 sive efforts have been made for analytical modeling of the non-12 linear interaction in SMFs using perturbative approaches [14, 13 15, 28–38]. In [14, 15], an analytical model based on Volterra 14 series transfer function (VSTF) has been developed to address 15 the nonlinear impairments in long-haul transmission. One of 16 these approaches is the well-known Gaussian noise model (GN-17 model), which is considered a reasonable-simple tool for ad-18 dressing the nonlinearity [38]. The GN-model concept has been 19 proposed in [36]. Then, it has been validated over a wide range 20 of SMFs systems [37-41]. Recently, several extension efforts 21 have been done on the GN model to enhance its accuracy and 2.2 consider new features such as the impact of modulation for-23 mats and stimulated Raman scattering on long-haul transmis-24 sion performance [42-46]. In this work, we discuss the non-25 26 linear Kerr-effects for transmission over FMFs. For simplicity, 27 we do not take into our consideration the impact of the mod-28 ulation techniques and other nonlinear impairments. The GN 29 model is extended for FMFs transmission, since the nonlinear 30 interference between two orthogonal-polarized fields is equ. 73-31 lent to that between two co-propagating spatial modes [47]. In 32 recent years, some numerical and analytical efforts been 33 developed for evaluating this nonlinear propagatic 1 in FN 7s 34 [10, 16, 48–54]. In [50], an analytical analysis of the onlin ar 35 interference in a weak coupled two-mode MDN based sy, tem 36 has been introduced. In [4, 10, 51], the applic tion of the GN-37 model in multi-mode fibers (MMFs) based syster is has been 38 validated. Furthermore, a generic expression for estimating the 39 nonlinear information spectral density of MMF. hased system 40 has been proposed in [52]. In our previou, work, we have just 41 presented simple closed-form express ons for the nonlinear in-42 terference power for both weak- and succe coupling regimes 43 44 [53, 54].

The main contributions in this parter are summarized as follows.

- A complete mathematical analy is has been explored for obtaining closed-form $e_{\lambda_1} = e_{\lambda_2}$ ons of the nonlinear interference power for ' oth we k- and strong coupling regimes over FMFs based ransmis ion systems.
 - The effect of **CAGD** and its management in week coupling transmiss. In .s discussed.
- Expressions for the nonlinearity accumulation through multiple spans fiber propagation is presented.
- Analytical results illustrate the impact of nonlinearity on the bit-error rate (BER) performance under different sys-

tem parameters, where the effect of both intra- and intermodal nonlinearities are discussed.

- The impact of linear mod coupling is analytically explored.
- The impact of the non'.net 'ity on the maximum reach of different optical fibe: sche.nes is discussed.

The remaining of this potents organized as follows. In Section 2, we review file WM in FMFs transmission and explore the derivation of sinplified expressions for the phasematching condition of a M process for both weak- and strongcoupling regimes in a ponimear propagation equations are also reviewed and the performince parameters are presented in same Section. In Section 3. rigorous mathematical derivation for the modifie i GN-model in FMFs is detailed in order to obtain closed-former propagating over multiple-spans fiber is discus. d. In Section 4, our results of the derived expressions are discussed and compared to similar cases in literature. Finally, we live the conclusions in Section 5.

Ceneral Considerations

2.¹. Four-Wave-Mixing in FMFs

The different nonlinear Kerr-effects that originate in FMFs transmission are summarized in Fig. 1. Self-phase modulation (SPM), cross-phase modulation (XPM), and four wave mixing (FWM) are third-order parametric processes that modulate the fiber refractive index [55, 56]. These nonlinear effects can be classified into: (a) intra- and inter-channel nonlinearity based on the frequency channel interactions and (b) intra- and intermodal nonlinearity based on space (mode) interactions [57, 58]. Both SPM and XPM processes can be treated as special types of the FWM process [36]. For a FWM process in FMFs, the nonlinear interaction process among spatial fields at frequencies (f_r, f_s, f_k) results in an energy-transfer into an idler mode with a frequency (f_i) , where (i, s, r, k) are the frequencies indices [59, 60]. This FWM nonlinear-interaction occurrence among different spatial fields requires two conditions to be satisfied [51, 55, 59, 60]: (1) a frequency (wavelength)/mode conservation condition $[if_0]_p = [rf_0]_m - [sf_0]_q + [kf_0]_l$, and (2) a phase-matching condition $\Delta \beta_{mqlp}^{isrk}(f_0) = \beta_m(rf_0) - \beta_q(sf_0) + \beta_l(kf_0) - \beta_p(if_0)$, where f_0 is the frequency separation between any two successive frequency-components and the subscripts (m, q, l, p) are the spatial modes indices.

Phase-Matching Condition

Simplified expressions for the phase-matching condition in FMFs can easily be formulated. For the weak coupling regime, the dispersion term of the *p*th mode can be expanded using Taylor's series as: $\beta_p(f) = \beta_{0_p} + 2\pi f \beta_{1_p} + 2\pi^2 f^2 \beta_{2_p} + \cdots$, where β_{0_p} , β_{1_p} , and β_{2_p} are the propagation constant, the group delay (GD) parameter, and the group velocity dispersion (GVD) parameter, respectively. We focus our consideration to significant terms

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Figure 1: Different nonlinear Kerr-effects in FMFs based system; SM-SPM: self-mode self-phase modulation, SM-XPM: self-mode cross-phase modulation, SM-FWM: self-mode four wave mixing, XM-XPM: cross-mode crossphase modulation, XM-FWM: cross-mode four wave mixing. (r, s, k, i) and (m, q, l, p) are the frequencies and spatial modes indices, respectively.

only (that is, up to the GVD term) [27, 61]. By substituting this expansion in the phase-matching condition, we obtain

$$\Delta \beta_{mqlp}^{ssrk}(f_0) = (\beta_{0m} - \beta_{0q} + \beta_{0l} - \beta_{0p}) + 2\pi [rf_o\beta_{1m} - sf_0\beta_{1q} + kf_0\beta_{1l} - if_0\beta_{1p}] + 2\pi^2 \{ [rf_o]^2\beta_{2m} - [sf_0]^2\beta_{2q} + [kf_0]^2\beta_{2l} - [if_0]^2\beta_{2p} \}.$$
(1)

In birefringent-fiber transmission, both SPM and XPM effects are dominant compared to the FWM effect [16, 47], thus th phase-matching conditions for both XPM and SPM (i.e., m = q, and l = p) can be rewritten as:

$$\Delta \beta_{ppqq}^{rski}(f_0) = 2\pi \left[(r-s)\beta_{1_q} - (i-k)\beta_{1_p} \right] f_o + 2\pi^2 \left\{ [(rf_o)^2 - (sf_0)^2]\beta_{2_q} + [(if_0)^2 - (kf_0)^2]\beta_{2_p} \right\}$$
(2)

Then, by applying the frequency conservation condu. v. i.e., $(rf_0 - sf_0 = if_0 - kf_0)$ and setting a simplified n tation for $\Delta \beta_{ppqq}^{isrk}(f_0)$ as $\Delta \beta_{pq}(f_0)$, we get:

$$\Delta \beta_{pq}(f_0) = 2\pi \left\{ (i-k) [\beta_{1_q} - \beta_{1_p}] f_o \right\} + \gamma^2 \left\{ [(rf_o)^2 - (sf_0)^2] \beta_{2_q} + [(if_0)^2 - (kf_0)^2] \beta_{2_p} \right\}.$$
 (3)

In order to obtain a continuous frequency domain expression, we substitute: $if_0 \to f$, $kf_0 \to f_1$, $if_0 \to f_2$, $sf_0 \to f_1 + f_2 - f$. This yields

$$\Delta \beta_{pq}(f) = 2\pi (f_1 - f) \Delta \beta_{1_{pq}} + 2\pi^2 (f_1 - f) [(f_1 - f + 2f_2)\beta_{2_q} - (f_1 + f)\beta_{2_p}],$$
(4)

where $\Delta\beta_{1_{pq}} = \beta_{1_q} - \beta_{1_p}$ is the *D* between spatial modes with indices p and q. At the cert or channel (i.e., i = 0), the last expression reduces to:

$$\Delta \beta_{j} = \gamma - f_{\star} (\Delta \beta_{1_{pq}} + 2\pi f_2 \beta_{2_q}). \tag{5}$$

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Moreover, for the intr. nodal case (i.e., q = p), the phasematching condition is further simplified to: $\Delta\beta_{pp} \approx 4\pi^2 f_1 f_2 \beta_{2p}$. Whereas, for strong coupling regime, the phase-matching condition is given by: $\Delta\beta \approx 4\pi^2 f_1 f_2 \hat{\beta}$, where $\hat{\beta}$ is the average GVD parameter of the co-propagating spatial modes for the strong coupling regime.

2.2. Signal Propagation in FMFs

The frequency-domain electric-field propagating in FMFs can be expressed for the *p*th mod ϵ [16] as:

$$\bar{\mathbf{E}}_p(x, y, z, f) = \mathcal{F}_p(x, y) \bar{\mathbf{A}}_n(z, f),$$
(6)

where $F_p(x, y)$ is the spati 1 field distribution and $\bar{\mathbf{A}}_p(z, f)$ is the slowly-varying field enve. e vector for the *p*th mode as in [27]. According to [16], the confinear propagation in nonlineardispersive FMFs can be described by the generalized multimode coupled Mana' ov e juai in for the pth mode as follows

$$\frac{\partial \bar{\mathbf{A}}_p(\bar{z}, f)}{z} = \bar{\mathcal{L}}_p(f) \bar{\mathbf{A}}_p(z, f) + \bar{\mathcal{G}}_{nl_p}(z, f).$$
(7)

The right-har $\frac{1}{3}$ side $\frac{1}{\sqrt{2}}$ is divided into two terms: the first one is the linear part we re $\overline{\mathcal{L}}_{p}(f)$ is a linear operator that includes both attenuation and dispersion operators, and the second term represents . • source of nonlinear interference due to the Kerreffects. Both $\lambda_{p}(f)$ and $\overline{G}_{nl_{p}}(z, f)$ are expressed based on the general. A multi-mode coupled Manakov equations for both coupling regimes in the following subsections (2.2.1 and 2.2.2). line. coupling between the co-propagating fields. However, the randomly-birefringence process averages out this linear part [1], 19].

??.1. Weak Coupling Regime

In this regime, the linear operator is expressed as $\bar{\mathcal{L}}_p(z, f) =$ $-\bar{\boldsymbol{\alpha}}_p - j\bar{\boldsymbol{\beta}}_p(f)$, where $\bar{\boldsymbol{\alpha}}_p$ and $\bar{\boldsymbol{\beta}}_p(f)$ are the fiber attenuation and dispersion operator for a dual-polarized field on the pth mode, respectively. Furthermore, the nonlinear term $\mathcal{G}_{nl_n}(z, f)$ is expressed as [16]

$$\bar{\mathcal{G}}_{nl_p}(z,f) = \frac{1}{3} \frac{4}{3} \gamma \sum_{q}^{M} f_{pq} \left(\frac{2}{3}\right)^{\delta_{pq}} \bar{\mathbf{A}}_q(z,f) * \left[\bar{\mathbf{A}}_q^{\star}(z,f)\right]^T * \bar{\mathbf{A}}_p(z,f).$$
(8)

Here, $\gamma = 2\pi n_2 / (\lambda A_{\text{eff}})$ is the fiber nonlinearity coefficient (with n_2 being the nonlinear-index coefficient, λ the propagating wavelength, and $A_{\rm eff}$ the core effective area of the fundamental mode), M is the number of the co-propagating modes, and δ_{pq} is the Kronecker delta function. The operator * donates the convolution, and the superscripts T and \star donate for the transpose and conjugation operators, respectively. The nonlinear interaction tensor $f_{pq} = f_{ppqq}$ between spatial modes with indices p and q is given by:

$$f_{ppqq} \stackrel{\text{def}}{=} A_{\text{eff}} \frac{\iint |F_p(x,y)|^2 |F_q(x,y)|^2 dxdy}{\iint |F_p(x,y)|^2 dxdy \cdot \iint |F_q(x,y)|^2 dxdy}.$$
 (9)

Note that the source of nonlinear interference part is classified into two distinct source-limited cases in FMFs; intramodal (self-mode modulation, SMM) and intermodal (cross-mode modulation, XMM) nonlinearity.

2.2.2. Strong Coupling Regime 1

In strong coupling regime, the linear operator is expressed as $\mathcal{L}_p(f) = -\hat{\alpha} - j\hat{\beta}(f)$, where $\hat{\alpha}$ and $\hat{\beta}$ are the attenuation coef-3 ficient and average GVD paramter of the co-propagating spatial modes, respectively. The nonlinear term in (7) is expressed as [8, 16]:

$$\bar{\mathcal{G}}_{nl_p}(z,f) = j\gamma\kappa \sum_{q}^{M} \bar{\mathbf{A}}_q(z,f) * \left[\bar{\mathbf{A}}_q^{\star}(z,f)\right]^T * \bar{\mathbf{A}}_p(z,f), \quad (10)$$

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where
$$\kappa = \sum_{\substack{p,q \in \{1,2,\dots,M\}:\\q \le p}} \frac{32}{2^{\delta_{pq}}} \frac{f_{pq}}{6M(2M+1)}.$$
 (11)

15 2.3. Performance Parameters

To assess the performance of optical communication systems, it is essential to evaluate the bit-error rate (BER). It normally depends on the modulation format characteristics and is obtained in terms of the signal-to-noise ratio (SNR) based on the modulation techniques and its constellation cardinality [62–64]. Moreover, the SNR for the *p*th mode propagating in multiple-spans of FMFs can be expressed as [36]:

$$SNR_p = \frac{B_n}{B_{ch}} \frac{P_{tx}}{P_n^{ac} + P_{nl_p}^{ac}},$$
(12)

27 where B_n is the noise bandwidth, B_{ch} is the channel bandwidt 28 P_{tx} is the average lunch power per mode, and $P_{nl_p}^{ac}$ is the accu-29 mulated nonlinear interference power per mode, to be de 30 in Section 3. P_n^{ac} is the accumulated complex optical amplific. 31 noise variance. For erbium-doped fiber amplifiers (EDFAs), it 32 is simply the amplified-spontaneous-emission (ASF noise per 33 lumped amplifier, and is expressed as: $P_{ASE} \approx (G - 1)Fh\nu_{I_n}$, 34 35 where G is the amplifier gain, F is the amplifie noise correction, re, 36 h is Plank's constant, and v is the center-ch anel frequency 37 [33, 38, 65]. 38

3. Modified GN-Model for Mode-Division multiplexing Sys-40 tems 41

42 In dual-polarized transmission over ^c MFs, the GN-model 43 treats the nonlinearity as an indeper dent additive Gaussian noise 44 source, which is statistically in over Jent from both the am-45 plifier noise and the transmitted sign 1 [18, 61]. This can be 46 applied to FMFs as the nor mear interaction among the co-47 propagating spatial (mode) fit 'ds is eq ivalent to that among the 48 49 orthogonal-polarized fields [47]. In the following subsections, 50 we explore the modelin (assur. ptions of the propagating sig-51 nal, followed by detailed derivations of expressions for the non-52 linear interference powers ... Joth strong- and weak-coupling 53 regimes. 54

55 3.1. Modeling Assumptions 56

In order to apply a perturbation analysis such as the GN-57 model, some assumptions should be considered for the trans-58 mitted signal [38, 51, 52]: (1) the signal Gaussianity, (2) the 59 statistical independence of the nonlinear interference from both 60

the ASE noise and the transmitted signal, (3) the mode dependent loss is negligible, and (4) the relative low to moderate level of nonlinearity. A complex periodic process, which is spectrally shaped to satisfy the above assu 'ptions, is used as a transmitted field envelope process at the input f the optical fiber (z = 0) using Karhunen-Loéve form¹, [62]:

$$A_{p_{x}}(0,f) = H_{p_{x}}(f) \sqrt{\sum_{v=-\infty}^{\infty} \vartheta_{v,p_{x}}} \,\delta(f - vf_{0}), \qquad (13)$$

where $H_{p_x}(f)$ is the t ansr tring filter shape, and ϑ_{v,p_x} is a random variable of the *th* mode on *x*-polarization at frequency (vf_0) having a zero mean. $F\{\vartheta_{v,p_x}\} = 0$ and a unity variance $E\{|\vartheta_{v,p_x}|^2\} = 1$, such that E is the expectation operation.

We aim at obtaining an a lalytical model for the nonlinearity in FMFs, thus y should obtain a closed-form solution of (7) as:

$$\bar{\mathbf{A}}_{p}(z,f) = e^{\hat{\mathcal{L}}_{p}z} \bar{\mathbf{A}}_{p}(0,f) + e^{\bar{\mathcal{L}}_{p}z} \int_{0}^{z} e^{-\bar{\mathcal{L}}_{p}z'} \bar{\mathcal{G}}_{nl}(z',f) dz', \quad (14)$$

where $\bar{\mathbf{A}}_{p}(f)$ is the transmitted optical field envelope vector. A strai thto, ver a linear solution, $A_{l_{n_x}}(z, f)$, of (7) can be obtained by subst. uting (13) in the linear part of (14) as:

$$A_{l_{p_x}}(z,f) = H_{p_x}(f) \ e^{\mathcal{L}_{p_x} z} \sqrt{f_0} \sum_{\nu=-\infty}^{\infty} \vartheta_{\nu,p_x} \delta(f-\nu f_0).$$
(15)

Uviously, obtaining the second part of the right hand side in (14) represents a big dilemma, because the source of nonlinear interference $\bar{\mathcal{G}}_{nl}(z, f)$ is a function of both the linear and nonlinear solutions. Moreover, the nonlinear solution depends on the source of the nonlinear interference. Fortunately, the linear solution (15) can be used as a perturbative start for obtaining the nonlinear-solution part in (14) through the GN-model scenario, as will be shown in the following two subsections.

3.2. Nonlinear Interference in Weak Coupling Regime

In this subsection, we follow a similar procedure as in [33, 37, 51] to obtain an expression for the nonlinear interference power, $P_{nl_n}^{w}$, in the weak coupling regime. We substitute the linear solution from (15) in the source of nonlinear interference $\overline{G}_{nl}(z, f)$ given by (8). We assume that all dual-polarized transmitting filter shapes are identical, i.e., $H_{p_x}(f_v) = H_{p_y}(f_v) =$ $H_{q_v}(f_v) = H_{q_x}(f_v) = H(f_v)$. Then, we perform the triple convolution operation and apply the frequency condition: $(sf_0 - rf_0 +$ $kf_0 = if_0$). The expression of $\overline{\mathcal{G}}_{nl}(z, f)$ for a particular mode at a specific frequency on x-polarization is obtained as:

$$\mathcal{G}_{nl_{p_{x}}}^{w.}(z,f_{0}) = j \frac{4}{3} \gamma f_{0}^{\frac{3}{2}} e^{-3\alpha z} \sum_{q}^{M} f_{pq} \left(\frac{2}{3}\right)^{\delta_{pq}} \sum_{i=-\infty}^{\infty} \delta(f-if_{0})$$

$$\times \sum_{r,s,k} \left[H(rf_{0})H^{\star}(sf_{0})H(kf_{0})\right] \vartheta_{k,p_{x}} \left(\vartheta_{r,q_{x}}\vartheta_{s,q_{x}}^{\star} + \vartheta_{r,q_{y}}\vartheta_{s,q_{y}}^{\star}\right) e^{-j[\beta_{q}(rf_{0})-\beta_{q}(sf_{0})+\beta_{p}(kf_{0})]z}, \quad (16)$$

where the superscript w. denotes the weak coupling regime. We use (16) as a perturbative start to solve the dilemma of obtaining the second term of (14). By recalling the linear operator expression $\mathcal{L}_p(f)$ and substituting by $[\beta_q(rf_0) - \beta_q(sf_0) + \beta_p(kf_0) - \beta_q(sf_0)]$

 $\beta_p(if_0) = \Delta \beta_{pq}(f_0)$], we get the nonlinear optical field solution as:

$$\begin{aligned} A_{nl_{p_{x}}}^{w.}(z,f_{0}) &= j\frac{4}{3}\gamma f_{0}^{\frac{3}{2}}e^{-\alpha z}\sum_{q}^{M}f_{pq}\left(\frac{2}{3}\right)^{\delta^{pq}}\sum_{i=-\infty}^{\infty}e^{-j\beta_{p}(if_{0})}\delta(f-if_{0}) \\ &\times \sum_{r,s,k}\left[H(rf_{0})H^{\star}(sf_{0})H(kf_{0})\right]\left(\vartheta_{r,q_{x}}\vartheta_{s,q_{x}}^{\star}+\vartheta_{r,q_{y}}\vartheta_{s,q_{y}}^{\star}\right) \\ &\times \vartheta_{k,p_{x}}\left[\frac{1-e^{-[2\alpha-j\Delta\beta_{pq}(f_{0})]z}}{2\alpha-j\Delta\beta_{pq}(f_{0})}\right]. \end{aligned}$$

$$(17)$$

The power spectral density, $S_{nl_{p_x}}^{w.}(z, f_0)$, of the nonlinear interference can be obtained by statistically averaging the square absolute value of the nonlinear optical field $E\{A_{nl_p}^{w.}(z, f_0)A_{nl_p}^{w.*}(z, f_0)\}$ as:

$$S_{nl_{p_{x}}}^{w.}(z,f_{0}) = \frac{16}{9}\gamma f_{0}^{2} e^{-2\alpha z} \sum_{q}^{M} f_{pq}^{2} \left(\frac{4}{9}\right)^{\delta_{pq}} \sum_{i=-\infty}^{\infty} \delta(f-if_{0})$$
$$\times \sum_{r,s,k} \left[H(rf_{0})H^{\star}(sf_{0})H(kf_{0})\right]^{2} E_{\vartheta\vartheta\star} \eta_{\text{FWM}}^{r,s,k}(f_{0}), \quad (18)$$

where $\eta_{\text{FWM}}^{rsk}(f_0) = |(1 - e^{-[2\alpha - j\Delta\beta_{pq}(f_0)]L_s})/(2\alpha - j\Delta\beta_{pq}(f_0)|^2)$ is the FWM efficiency [66], and $E_{\vartheta\vartheta\star}$ is expressed as $E\{(\vartheta_{r,q_x}\vartheta_{s,q_x}^{\star}\vartheta_{k,p_x} + \vartheta_{r,q_y}\vartheta_{s,q_y}^{\star}\vartheta_{k,p_x})(\vartheta_{r',q'_x}^{\star}\vartheta_{s',q'_x}^{\star}\vartheta_{k',p'_x} + \vartheta_{r',q'_y}^{\star}\vartheta_{r',q'_y}\vartheta_{k',p'_x})\}$.

The value of this expectation is altered for different norlinearity limits. For intermodal nonlinearity limit (XMM), by recalling the random variable's properties at the state: {(a' = q, p = p') and (r = r', s = s', k = k')}, the value of $E_{\theta\theta^*}$ eq. 71s "2". For intramodal nonlinearity limit (SMM), the averaging operation is performed at the aforementioned state is the intermodal limit besides an additional new state: {(r = p') and (s = k', r = r', k = s')}. This new state produces an additional "1" that makes the overall value of $E_{\theta\theta^*}$ equal, "3". So, the expression of the power spectral density (PSD) can' erev ritten as follows

$$S_{nl_{p_{x}}}^{w.}(z,f_{0}) = \frac{32}{9} \gamma^{2} f_{0}^{3} e^{-2\alpha z} \sum_{q}^{M} f_{pq}^{2} \left(\frac{2}{3}\right)^{\delta_{p}} \sum_{r=-\infty}^{\infty} \delta(j - if_{0})$$
$$\times \sum_{r,s,k} \left[H(rf_{0}) H^{\star}(s,f_{0}) h_{\chi}^{-r}(\dot{r}_{0}) \right]^{2} \eta_{\text{FWM}}^{r,s,k}(f_{0}).$$
(19)

The same expression for y-polarize on effect is obtained by performing similar analysis. The nonlinearity is evaluated at the span end where the ampli ier complements for the span loss. Thus, the overall PSD, i.e., $S_{nl_p}^{+}(f_0) + S_{nl_{p_x}}^{w}(f_0) + S_{nl_{p_y}}^{w}(f_0)$ can be expressed by

$$S_{nl_{p}}^{w}(f_{0}) = \frac{64}{9} \gamma^{2} r^{3} \sum_{h} f_{pq} \left(\frac{\gamma}{3}\right)^{\delta^{pq}} \sum_{i=-\infty}^{\infty} \delta(f - if_{0})$$
$$\times \sum_{r,s,k} \left[H(f_{0}) H^{\star}(sf_{0}) H(kf_{0}) \right]^{2} \eta_{\text{FWM}}^{r,s,k}(f_{0}).$$
(20)

The transmitting filter shape is assumed to be flat over the channel bandwidth, such that, $H_p(f_v) = (P_{tx}/2B_{ch})^{0.5} \operatorname{rect}(f_v)$ [61]. Furthermore, the discrete summation in (20) can be converted into continuous integral by setting $S_{nl_p}^{w}(f) = \lim_{f_0 \to 0} S_{nl_p}^{w}(f_0)$. Thus, the PSD expression can be expressed as follows

$$S_{nl_p}^{w.}(f) = \frac{8\gamma^2}{9} \frac{P_{tx}^3}{B_{ch}^3} \sum_{q}^{M} f_{pq}^2 \left(\frac{2}{3}\right)^{\delta} \int_{D} \int_{D} \eta_{\text{WM}}(f_1, f_2) df_1 df_2.$$
(21)

Here *D* is the spectral integration area, shown as the dark-gray area in Fig. 2a and $\eta_{\text{FWM}}(j_1, f_2) = \lim_{f_0 \to 0} \eta_{\text{FWM}}^{r,s,k}(f_0)$ is the FWM efficiency. $\eta_{\text{FWM}}(f_1, f_2) \approx 1$ be expanded using the phase-matching condition in (5) and und *i* the condition, $(\Delta\beta_{pq} \ll 2\alpha)$, at the center channel as:

$$\eta_{\text{FWM}}(f_1, f_2) \approx \frac{L_{\text{eff}}^2}{1 - \sum_{\text{eff}, a}^2 \left[2\pi f_1(\Delta\beta_{1_{pq}}\delta_{pq} + 2\pi f_2\beta_{2q})\right]^2}, \quad (22)$$

where $L_{\text{eff}} = (1 - e^{-\alpha L_s})/2\alpha$ and $L_{\text{eff},a} = 1/2\alpha$ are the effective and asymphtic-enective lengths of a fiber with a span length L_s , respectively [38]. We use the approximation of the spectral bands in γ a square integration area, shown as light-gray area in Fig. 2a. This spectral approximation is verified to give a characteristic to the exact integral evaluation [67]. Furthermore, this poroximation reduces the over-estimation of the nonlinear interference power in the GN-model. An analytical expression of ne PSD, $S_{nl_p}^{w}$, formulated by integrating the FWM efficiency in (22) over the light-gray area in Fig. 2a using the integration γ^{1} -ntities in [68]. Then, by integrating the obtained analytical expression of the PSD over the noise bandwidth B_n , a closedform expression for the per-span nonlinear interference power, $P_{nl_s}^{w}$, can be obtained as:

$$P_{nl_p}^{w.} \approx \frac{4}{9\pi} \gamma^2 \frac{L_{\text{eff}}^2}{L_{\text{eff},a}} \frac{B_n}{B_{ch}^3} P_{tx}^3$$
$$\times \sum_{q}^{M} \frac{f_{pq}^2}{3^{\delta_{pq}} |\beta_{2_q}|} \left[\operatorname{arcsinh}(\psi^+) + \operatorname{arcsinh}(\psi^-) \right], \quad (23)$$

where $\psi^{\pm} = \frac{\sqrt{3}}{4}\pi L_{\text{eff},a}B_{\omega}(\frac{\sqrt{3}}{2}\pi|\beta_{2_q}|B_{\omega} \pm \Delta\beta_{1_{pq}})$. Here $B_{\omega} = B_{ch}N_{ch}$ is the total WDM bandwidth, and N_{ch} is the number of the WDM channels.

3.3. Nonlinear Interference in Strong Coupling Regime

By starting from the MM-NLSE for strong coupling regime (10) and the phase-matching condition, we apply the same procedure as has been explored above in Section 3.2. The estimated value of $E_{\theta\theta^*}$ equals "3". A closed-form expression for the per-span nonlinear interference power, $P_{nl_p}^{s}$, is thus obtained as:

$$P_{nl_p}^{s.} \approx \frac{3M}{8\pi} \frac{\gamma^2 \kappa^2}{|\hat{\beta}_2|} \frac{L_{\text{eff}}^2}{L_{\text{eff},a}} \frac{B_n}{B_{ch}^3} P_{tx}^3 \operatorname{arcsinh}\left(\frac{3}{8}\pi^2 L_{\text{eff},a} |\hat{\beta}_2| B_{\omega}^2\right), \quad (24)$$

the superscript s. denotes the strong coupling regime.



Figure 2: a) Spectral integration areas; D (dark-gray): the original integration area with limits $[-B_{\omega}/2, B_{\omega}/2]$ for all $(f_1, f_2, f_1 + f_2)$ and S (light-gray): the approximated square limits with a side length of $= \sqrt{3}B_{\omega}/2$, and B_{ω} is the total bandwidth, b) Spatial field distribution for the six LP spatial modes.

3.4. Accumulation of Nonlinear Interference over Multiple Spans

The accumulation scenarios of the nonlinear interference power through propagating over multiple-spans fiber can be 20 viewed as either a) coherent approach (accumulating nonlinear interference fields) [61], or b) non-coherent approach (accumu-22 lating nonlinear interference powers). The second approach can 23 be modified from a pure-linear variation with the number of spans N_s to a super-linear with an exponent ($\varepsilon < 1$), given by [69]:

$$\varepsilon \approx \begin{cases} \frac{3}{10} \ln\left(1 + \frac{12}{L_s} \frac{L_{\text{eff},a}}{\operatorname{arcsinh}(\psi^+) + \operatorname{arcsinh}(\psi^-)}\right); & \text{weak,} \\ \frac{3}{10} \ln\left(1 + \frac{6}{L_s} \frac{L_{\text{eff},a}}{\operatorname{arcsinh}(\frac{3}{8}\pi/\hat{\beta}|L_{\text{eff},a}B_{\omega}^2)}\right); & \text{strong.} \end{cases}$$
(25)

32 Thus, the total accumulated nonlinear interference power and total amplifier noise in (12) can be written as $P_{nl_s}^{ac} \cdot N_s^{1+c} \gamma_{u_p}$ and $P_n^{ac} \approx N_s^{1+\epsilon} P_{ASE}$, respectively.

4. Results and Discussions 37

In this section, we apply the modified G' model to a generic 39 long-haul hybrid wavelength-division multiplex. or and mode-40 division multiplexing (WDM-MDM) s on m with the param-41 eters similar to those in [16]. A st o-in ex few-mode fiber 42 (SI-FMF) is used as an optical channel . " the weak coupling 43 44 regime. This SI-FMF has a nume ical aperture of 0.2, a core 45 diameter of 12.5 μ m, and a norm. "ize 1 frequency of $V \approx 5$ at a 46 wavelength of 1.55 µm. It supports sull' learly-polarized (LP) 47 spatial modes (LP₀₁, LP_{11a}, LP₀₂, LP_{21a}, LP_{11b}, LP_{21b}). The 48 spatial distributions of these 'P mod's are shown in Fig. 2b 49 and their dispersion coefficient and MGD are given in Table 1 50 [16, 70]. The fiber atten lation c. efficient α of 0.22 dB/km and 51 the nonlinear coefficient $\sqrt{1.4} W^{-1} km^{-1}$ are the same for all 52 The calculated values of the nonlinco-propagating modes. 53 ear tensors $(f_{pq} = f_{pp, -})$ t in different LP modes are listed in 54 Table 2. In addition, w study a graded-index few-mode fiber 55 (GI-FMF) as the channel for the strong coupling regime. This 56 GI-FMF supports different Hermitian Gaussian (HG) modes 57 (HG₀₀, HG₀₁, HG₀₂+HG₂₀, HG_{11a}, HG₁₀, HG_{11b}) correspond-58 ing to the six LP modes of the SI-FMF [16]. The GI-FMF pa-59 rameters are; fiber attenuation α , dispersion D, and nonlinear γ 60

Table 1: Dispersion coefficient; D [ps/km · nm], differential mode group delay; DMGD [ns/km], and core effective areas $A_{\rm eff}$ [μm^2] for the six LP spatial modes.

	LP ₀₁	LP _{11a}	LP	LP _{21a}	LP _{11b}	LP _{21b}
D	25	27.3	-2.3	208	27.3	20.8
DMGD	0	6.5	2.	12	6.5	12
$A_{\rm eff}$	80	76	8?	86	76	86

Table 2: Calculated values $\int f_{npqq}$ for the six LP spatial modes.

	LP	лг ₀	LP_{11a} , LP_{11b}	LP_{21a}, LP_{21b}				
LP ₁₀	1.000	0.734	0.661	0.455				
LP_{02}	1.131	0.264	0.369	0.335				
LP_{11a} , LP_{11b}	0.660	0.369	1.053	0.608				
LP _{21a} , LP _{21b}	455	0.335	0.608	0.930				

coefficients of 0.22 $_{\mu}B/km$, 21.5 ps/km \cdot nm, and 1.4 W⁻¹km⁻¹, respectively. We consider dual polarized-multiplexing quadrature phase-shift keying modulation with $R_s = 28.5$ GBaud, that equ. 1s to . ". throughput of 25 GBaud and 14% of forward error connection (FEC) overhead. This corresponds to a WDM-. Inner bandwidth at the Nyquist border. EDFAs have 6 dB noise foure and a gain that compensates for the span loss, i.e., $2\alpha L_s$. The total fiber length is 1000 km with a span length Ū. ⁴ 100 km. These parameters are selected similar to those in [1,] in order to be able to compare their trends.

Figure 3 illustrates the effect of the different nonlinear penalies on the performance of FMFs based systems. We opt the standard single-mode fiber (SSMF) as a reference case study with the parameters: $\alpha = 0.22 \text{ dB/km}$, $D = 16.7 \text{ ps/km} \cdot \text{nm}$, and $\gamma = 1.3 \,\mathrm{W}^{-1} \mathrm{km}^{-1}$ [61]. Here, we discuss two transmission cases for each coupling regime in FMFs. The first one is called single-mode transmission (SMT) corresponding to turning on the fundamental mode (LP_{01}) only. The second case is called full-mode transmission (FMT) corresponding to turning on all the six co-propagating modes. The bit-error rate (BER) averaged over all the turning-on modes is depicted as a function of the average lunch power per mode. In linear region, increasing the average lunch power enhances the system performance. However, after the average lunch power reaches a specific level (optimal average lunch power), the nonlinear interference power becomes significant compared to the noise power level. Beyond this power, any increase in the lunch power leads to a degradation of the system performance. This optimal average lunch power per mode, that achieves the minimum BER (minimal points on curves) [71], can be formulated in weak coupling regime as

$$P_{tx_{opt}}^{w.} = \sqrt[3]{\frac{9\pi L_{eff,a} B_{ch}^{3} (G-1)Fh\nu}{4\gamma^{2} L_{eff}^{2} \sum_{q}^{M} \frac{f_{pq}^{2}}{3^{\delta pq} |\beta_{2q}|} \left[\operatorname{arcsinh}(\psi^{+}) + \operatorname{arcsinh}(\psi^{-}) \right]}}.$$
 (26)

It is proportional to the WDM-channel bandwidth and inversely to both the nonlinear tensors values and the number of copropagating modes M. Moreover, the existence of the DMGD increases this optimal average lunch power value. It does not

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Figure 3: Average BER versus average lunch power per mode for different transmission cases; B2B: back-to-back transmission (dashed-dotted), SSMF: standered single mode fiber (circles), SI-FMF: step-index few-mode fiber (squares), GI-FMF: graded-index few-mode fiber (diamonds), SMT: single-mode transmission (solid), FMT: full-mode transmission (dashed). The PM-QPSK/WDM-MDM system has $N_{ch} = 11$.

depend on the overall link length but on the fiber span length. Also, this power value is affected by the nonlinearity sourcelimited cases (SMM and XMM). Furthermore, in the strong coupling regime, the optimal average lunch power can be obtained as:

$$P_{tx_{opt}}^{s.} = \sqrt[3]{\frac{8\pi |\hat{\beta}_2| L_{eff,a} B_{ch}^3 (G-1) Fh\nu}{3M\gamma^2 \kappa^2 L_{eff}^2 \operatorname{arcsinh}\left(\frac{3}{8}\pi^2 L_{eff,a} |\hat{\beta}_2| B_{\omega}^2\right)}}.$$

For SMT case, the nonlinear interference resul s from he intramodal interaction. In GI-FMF based system (s. ong c/ apling regime), the intramodal nonlinear interference is mether than that in SSMF based system. Moreover, *i* SI MF based system (weak coupling regime), this intram dai . only lear interference is lower than that for both GI-FN G and SSMF based systems. This performance is due to different propagation properties, i.e., the dispersion coefficient of ... fundamental mode LP_{01} in SI-FMF based system is high r omp² red to the other two fiber schemes. This leads to reducing the inpact of nonlinearity in SI-FMF based system. In addition, the intramodal nonlinear penalty is altered with different f. or / fect ve areas for various fiber schemes. On the other hand, for $\sum \Gamma$ case, the nonlinear interference results from bot' intra- nd inter-modal nonlinear interactions. Thus, the diffe once in performance penalty between both FMT and SMT is and the intermodal nonlinear interference. For FMT ca e, it is . ticed that the GI-FMF based system suffers more that the SI- MF based system (taking the DMGD effect into account, this result is due to the impact of DMGD on the SFN r cased system, which reduces the effect of nonlinearity compared to that of GI-FMF based system which is theoretically considered a fiber scheme with zero-DMGD. Although, the randomly-averaging of the nonlinear interference in the GI-FMF based system (strong coupling regime) is greater than that in the SI-FMF based system (weak coupling regime), the DMGD effect in SI-FMF based system is dominant



Figure 4: Aver ge BFP versus average lunch power per mode in SI-FMF based system (weak cour .ng 1 gime), for different transmission cases; one [SMT] (diamonds), two (square), three (circles), and six [FMT] (trinagles) modes, B2B: back-to `ack transmission (dashed-dotted), for both unmangeded- (solid) and mangeded-D₁. GD (dashed) SI-FMF based system. The PM-QPSK/WDM-MDM $_{\odot}$ rem has $l_{ch} = 11$.

GI-r. F based system.

Figure 4 depicts the impact of DMGD and its management vr the performance of SI-FMF based system. The average B.'R is plotted versus the average lunch power per mode for "ferent number of co-propagating (turning-on) modes in SI-FMF (weak coupling regime). Two propagation systems are investigated: unmanaged-DMGD SI-FMF (with DMGD values as given in Table 1) and managed-DMGD SI-FMF based systems. It is shown that, for SMT case, the performance of both unmanaged- and managed-DMGD propagating systems are identical. For three co-propagating modes (LP₀₁, LP_{11a}, LP₀₂) case, the optimal performance of unmanaged-DMGD system is degraded by about "1" order of magnitude when compared with the managed-DMGD one. This can be explained as follows. In the managed-DMGD system, the effect of the low dispersion coefficient of the third mode, LP₀₂, leads to high nonlinear interference. While, in the unmanaged-DMGD one, the effect of DMGD of LP₀₂ compensates the effect of its low dispersion that reduces the overall nonlinear interference. For FMT case, the performance of unmanaged-DMGD system is better than that of three co-propagating modes. This performance can be explained as follows. By turning on the third mode (LP₀₂), its low dispersion coefficient results in higher nonlinear interference than that resulting form turning on one of the modes (LP_{11b}, LP_{21a}, LP_{21b}). Thus, the averaging over six modes results in an averaged BER value lower than that when turning on only three modes $(LP_{01}, LP_{11a}, LP_{02})$. On the other hand, the performance of managed-DMGD system is severely degraded when more than two modes are co-propagating, i.e., $M \in \{3, 4, 5, 6\}$. In addition, the average BER performance is approximately unchangeable. This performance is due to the low dispersion-coefficient of the mode (LP_{02}) when removing the DMGD effect. Specifically, for FMT case, the optimal performance of managed-DMGD system is degraded by about "1"



Figure 5: BER versus OSNR for different LP modes in a SI-FMF based system, including SMT: single-mode transmission (dashed), FMT: full-mode transmission (solid), and B2B: back-to-back transmission (dashed-dotted). The PM-QPSK/MDM system has $N_{ch} = 1$ and $P_{tx} = 4$ dBm.

order of magnitude compared with the unmanaged-DMGD system. In addition, the optimal power is reduced by about 1 dBm.
Thus, the DMGD-management increases the overall nonlinearity effect compared to the DMGD-ummanged based system.
Fig. 4 shows a potential agreement with the conclusions in [24].
It is worth mentioning that DMGD-management is required to reduce the receiver complexity to achieve more realistic MDM.
receiver [24, 25].

Figure 5 illustrates the nonlinear penalty due to both 11. 73and inter-modal nonlinear interactions on the BER performance for various LP modes, in SI-FMF (weak coupling regime) The BER is drawn versus the OSNR (optical signal-to noise ra io with respect to the ASE noise power with a reference of se bandwidth of 12.48 GHz (0.1 nm) [16]). Here, t' e single-n.ode transmission [SMT] is related to single transmission of 7 ly LP mode in a SI-FMF. For SMT case, the propa ating nating al field suffers from only the intramodal nonliner 'v. But, for FMT case, it suffers from both the intra- and inter-ni, ⁴al nonlinear interactions. Thus, the performance per an es between the two transmission cases is related to the oter odal nonlinear interaction. It is shown that all the mode (except LP_{02}) have almost the same intramodal non' near ty penalty. The LP₀₂ mode has a higher intramodal no... 'ne .r int rference because of its low dispersion coefficient. Moreo, r the non-degenerated modes (LP01 and LP02) have ower in ermodal nonlinear penalties compared with the degenerated ones (LP_{11v} and LP_{21v}). This is due to their deger the number that causes a high intermodal nonlinear penalt between $LP_{\nu\nu a}$ and $LP_{\nu\nu b}$ modes. In other words, these degen vated r odes are affected by the same propagation characteristics. Specifically, at the FEC-requirement (BER = 10^{-3}), the F. (T / asc suffers from an OSNR penalty of about 3 dB compared to the SMT case, for the $LP_{11a(b)}$ mode. This OSNR penalty is almost zero for the LP_{01} and LP_{02} modes.

Figure 6 illustrates the impact of linear coupling on the system performance between different number of co-propagating



Figure 6: Ave age BFP versus OSNR for different number of co-propgating modes in GI- MF or w ak- (solid), strong- (dashed) coupling regimes, and B2B: back-to-back trans aission (dashed-dotted). The PM-QPSK/MDM system has N_{ch} ¹ and $\tau_{tx} = 4$ dBm.

mous in usual coupling regimes, regardless of the effect of DMGD. The average BER is depicted versus OSNR for difte. int number of co-propagating modes for both weak- and strong Jupling in GI-FMF based system. Here we eliminate the μ_1 .GD effect in order to provide a fair comparison between t two distinct coupling regimes. Clearly, the performance of strong coupling regime is better than that of the weak coupling one in GI-FMF based system. The stronger linear mode coubling between the co-propagating modes, the higher variations in the propagating optical signal through the GI-FMF. This can be explained as follows. The nonlinear interference is reduced in the strong coupling GI-FMF compared to that in the weak coupling case. This nonlinearity reduction in the strong coupling regime is due to the higher randomly-averaging birefringence operation compared to the weak coupling case that lacks the high linear coupling effect. Furthermore, it is shown that the more number of co-propagating modes, the higher degradation in performance of the weak coupling regime compared to that of strong coupling one. Turning on more co-propagating modes increases the total linear mode coupling on a specific mode. This leads to an increase in the nonlinearity compensation because of the high randomly-averaging birefringence operation of the nonlinear interference power in the strong coupling regime compared to the weak coupling one. The analytical results of Figs. 5 and 6 follow same trends in [16]. Thus, it gives a window of verification for the GN-model in MDM based systems by comparing our results with those in [16].

Figure 7 shows the nonlinear penalty on the maximum distance that can be reached through various SI-FMF and GI-FMF for different co-propagation systems (SMT and FMT) at a BER of 10^{-3} . It is found that increasing the optical lunched power increases the achieved maximum reachable distance. But, after reaching a specific power the nonlinear interactions among copropagating modes, this nonlinear interference power becomes significant compared to the amplifier noise, and then an optimal maximum reach is achieved. The aforementioned contributions



Figure 7: Optical average lunch power versus fiber maximum reach for two distinct propagation systems: FMT (solid) and SMT (dashed) in different fibers; unmangeded-DMGD in SI-FMF (circles), mangeded-DMGD in SI-FMF (squares), and GI-FMF (diamonds). The PM-QPSK/WDM-MDM system has $N_{ch} = 11$.

of different fiber configurations are explored on the maximum reach. Specifically, the optimal maximum reach that could be achieved in the unmanaged-DMGD SI-FMF outreaches the GI-FMF case by about 777 km and 1100 km for both SMT and FMT, respectively. On the other hand, the managed-DMGD SI-FMF case reduces the optimal maximum reach by about 1350 km and 240 km compared to what can be achieved by the unmanaged-DMGD SI-FMF and GI-FMF systems, respectively.

5. Conclusions

The nonlinear interference penalty in birefrir gent few-mode fibers (FMFs) has been addressed by adapting the GN model for weak- and strong-coupling transmission f rough TV Fs based system. After a rigorous mathematical der v. ion, closed-form expressions for the nonlinear interference power 1. ve been derived. The nonlinearity accumulation and the DMGD effect through multiple-spans FMFs have 1 on onsidered. The results show that the nonlinear penalty becomes significant beyond an optimal average lunch pr wer nat is inversely proportional to the number of co-propagain modes. The unmanaged-DMGD weak coupling transm² such out rforms the strong coupling one due to the DMGD npact. In the other hand, regardless of the DMGD impact, the PER p rformance of strong coupling transmission is bet in than inat of the managed-DMGD weak coupling one. D 4GD m nagement increases the nonlinear penalty level and once he optimal power is reduced, which results in a de antion of the corresponding optimal system performance in L M' D managed based systems. Furthermore, the birefringence `ffect in weak coupling-based system is lower than that in strong coupling based one. Thus, an increase in the level of the linear mode coupling (i.e., turning-on more modes) leads to a higher reduction in the nonlinearity of the weak coupling-based system compared to the strong coupling based one. The same effects of the nonlinearity on the maximum reach are noticed.

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